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Burdick

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(54) **CABLE COMPENSATOR CIRCUIT FOR CCD VIDEO PROBE**

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(52) U.S. Cl. **333/28 R; 356/241.1; 600/109**

(58) Field of Search **333/28 R; 356/241.1; 600/109**

(56) **References Cited**

U.S. PATENT DOCUMENTS

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* cited by examiner

Primary Examiner—Robert Pascal

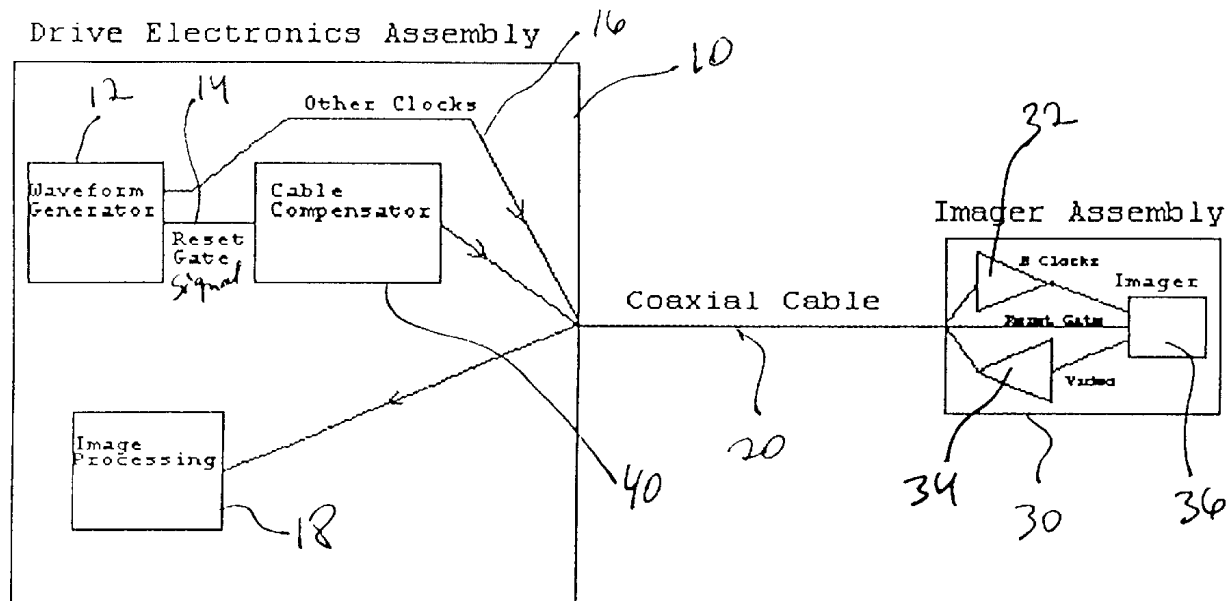
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(57) **ABSTRACT**

Picture noise in a CCD video probe is reduced by producing a faithful reproduction of a reset-gate signal received from a timing generator to an imager by using a cable compensator network. The network contains an RLC circuit which corrects the relative amplitude and phase of the various frequency components of the reset-gate signal. The precise values of the components in the RLC circuit are selected according to the attenuation and distortion characteristics of the coaxial cable connecting the drive electronics assembly from the imager assembly.

18 Claims, 5 Drawing Sheets



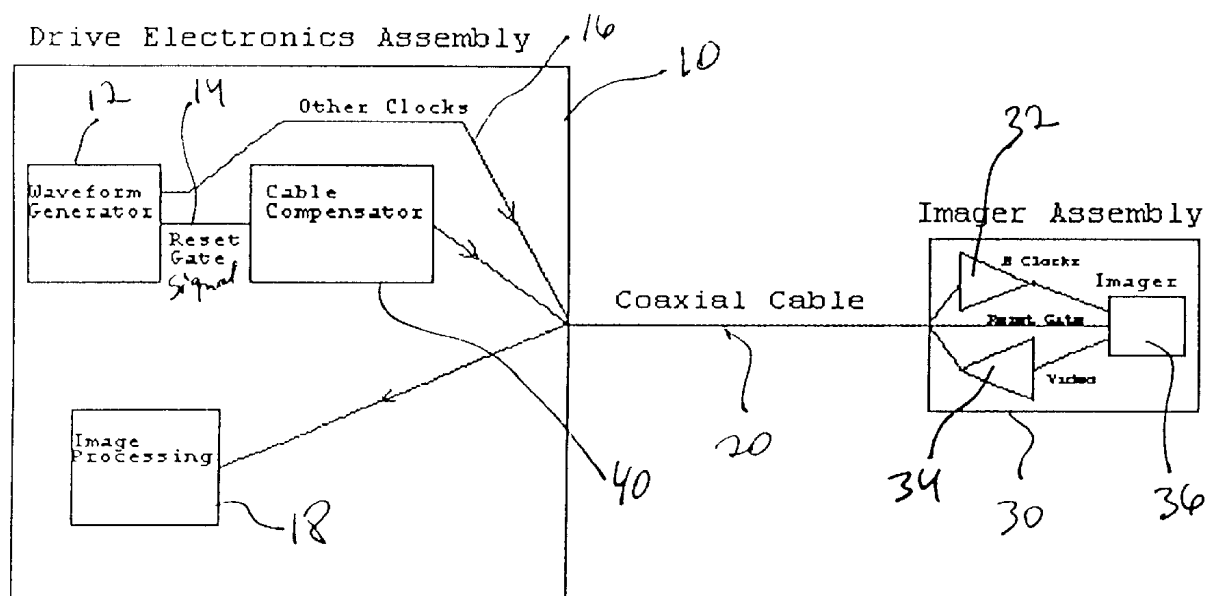


Fig. 1

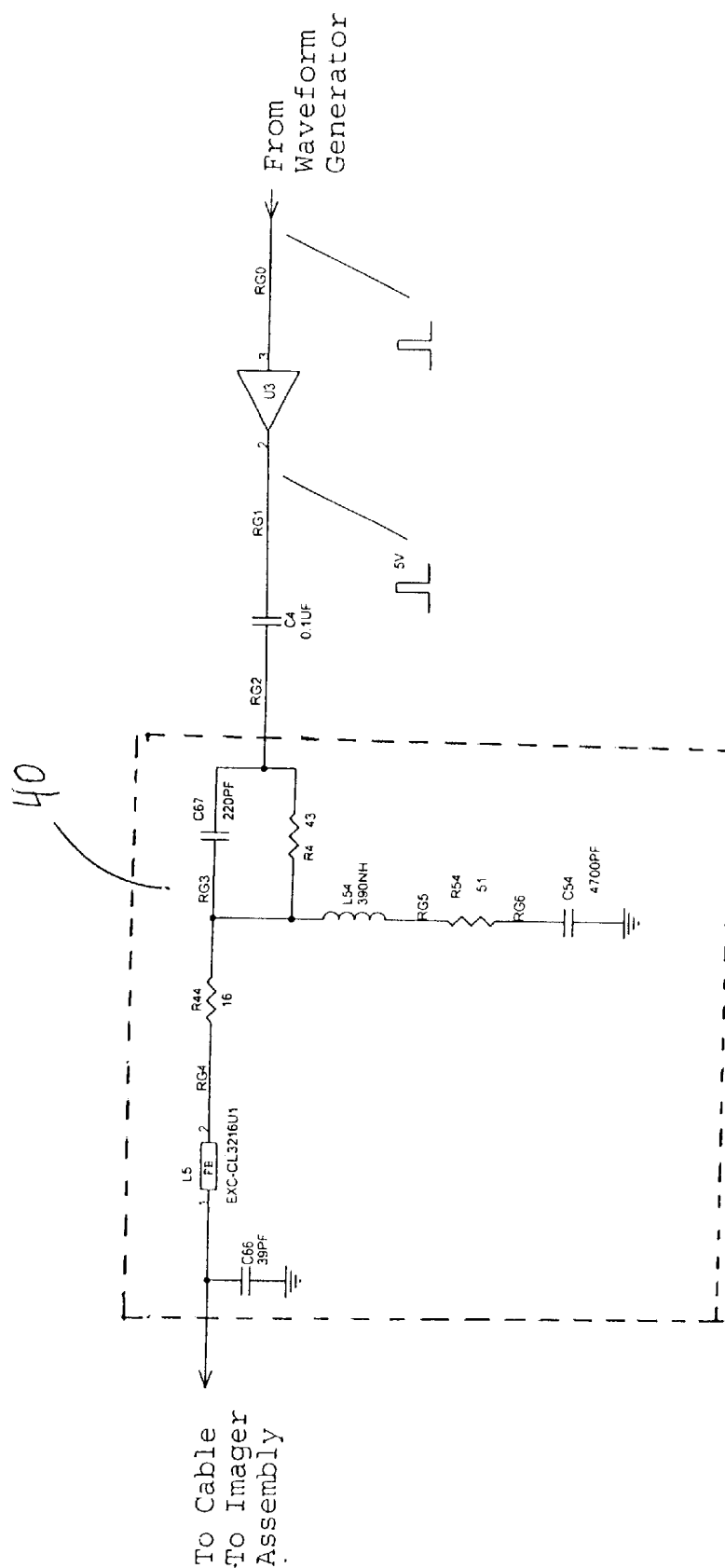


Fig. 2

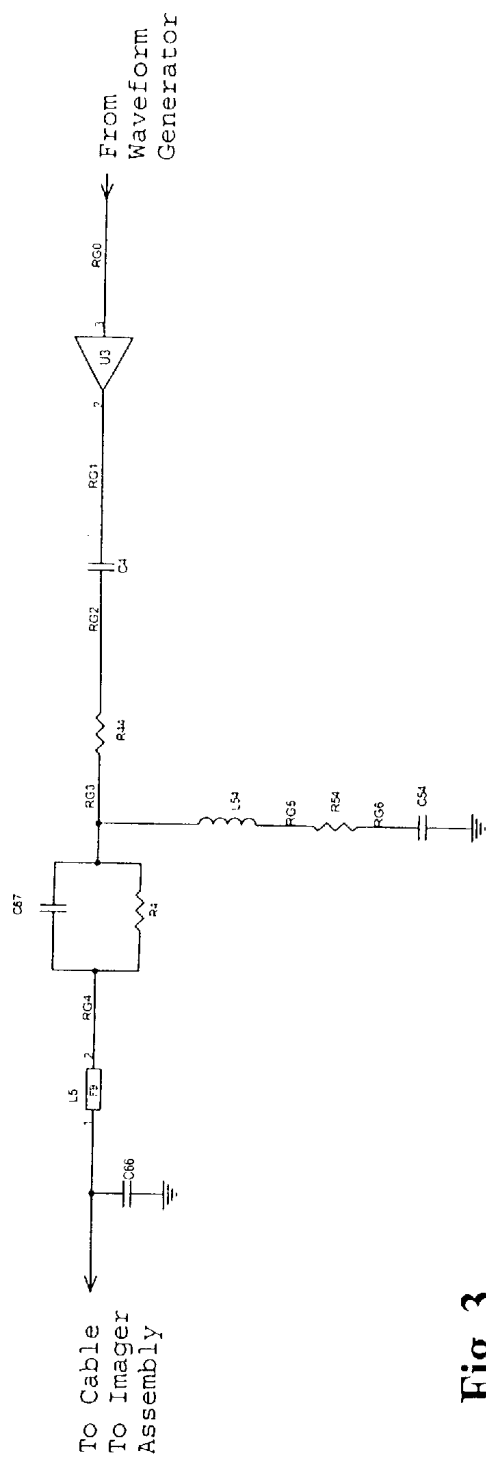


Fig. 3

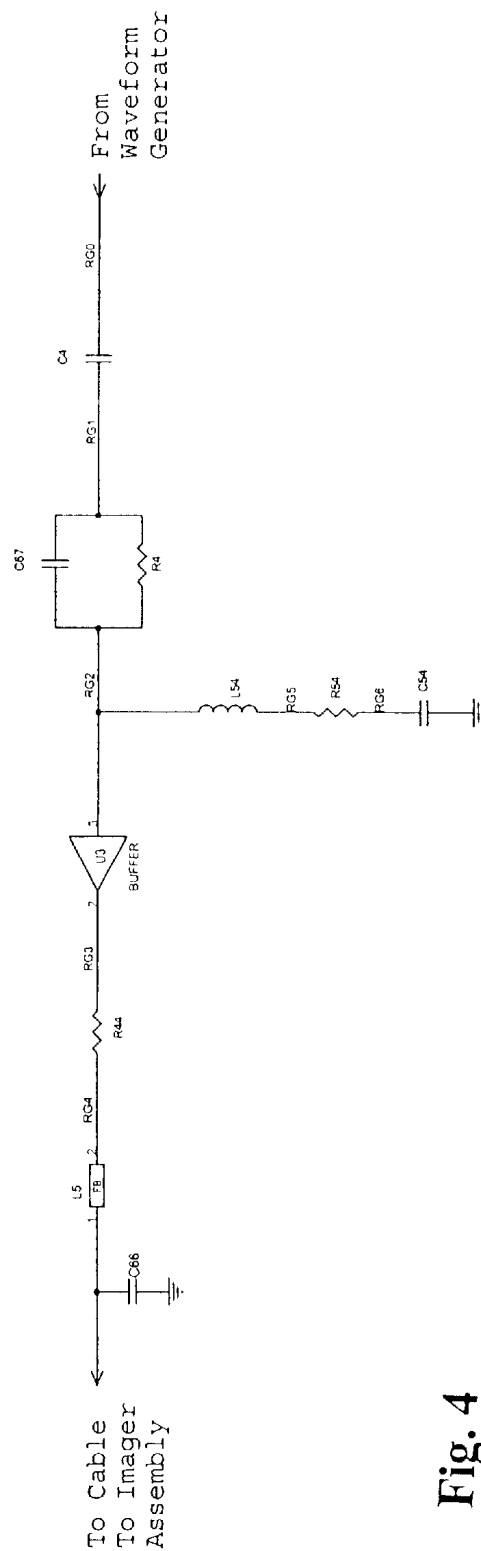
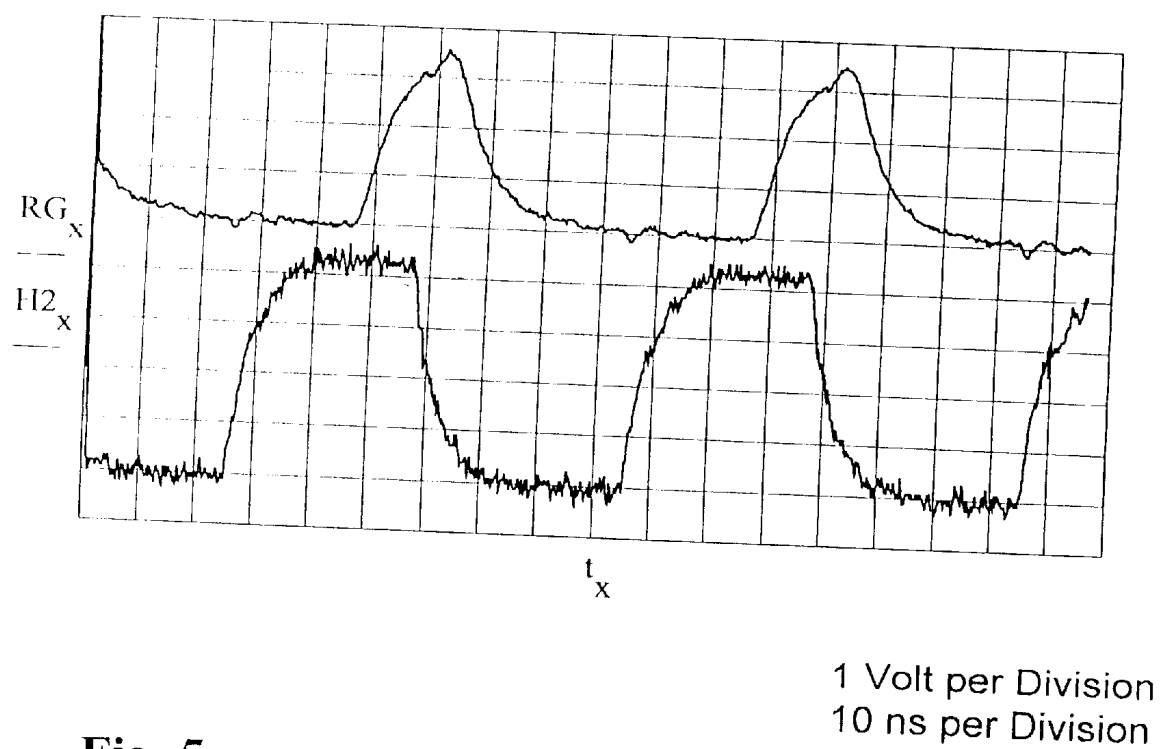


Fig. 4

**Fig. 5**

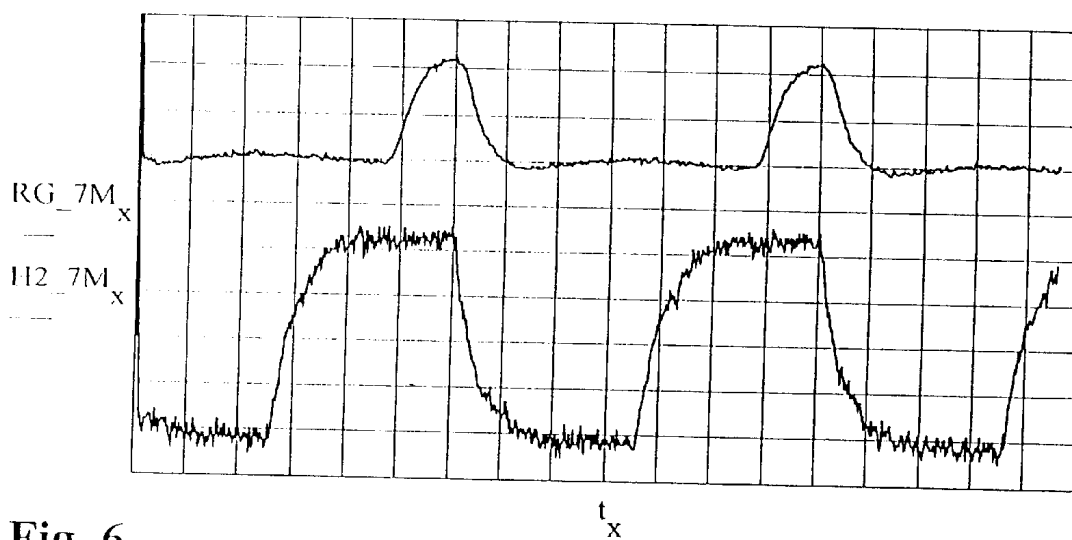


Fig. 6

1 Volt per Division
10 ns per Division

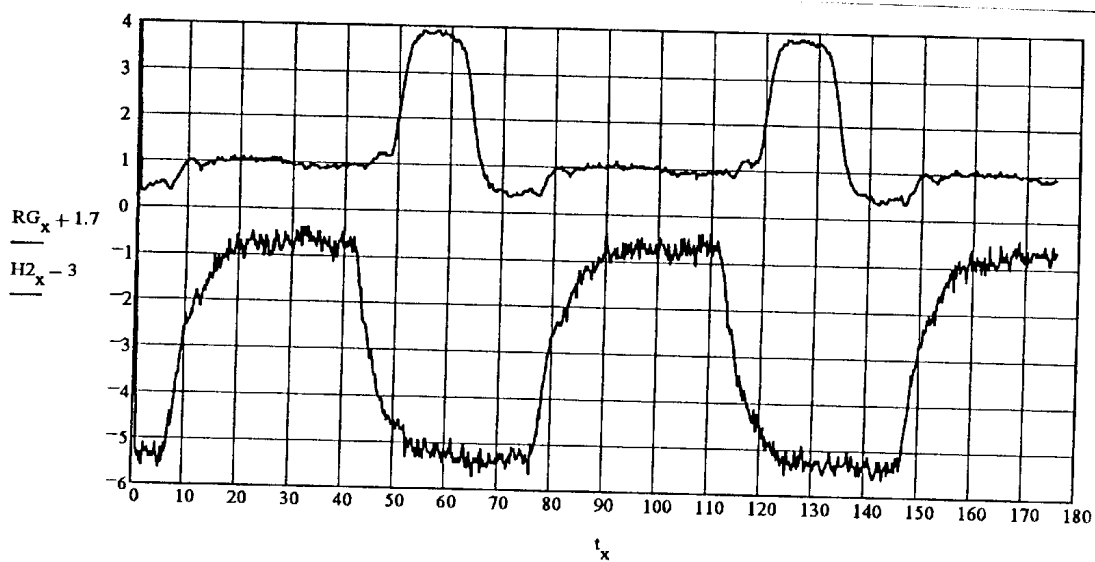


Fig. 7

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CABLE COMPENSATOR CIRCUIT FOR CCD VIDEO PROBE

FIELD OF THE INVENTION

The invention pertains to the field of CCD video probes, and in particular to long borescopes or similar medical devices that have long cables coupling the drive electronics assembly to the imager assembly.

BACKGROUND OF THE INVENTION

CCD (charge coupled device) imagers require the use of a reset-gate pulse to discard charge from previous pixels before each succeeding pixel is read out from the imager. This signal is both attenuated and distorted as it passes through a coaxial cable, with the effect being stronger for long, small diameter cables such as used in borescopes. The distortion of the reset-gate pulse causes it to spread out in time. This spreading and reduced amplitude cause both fixed-pattern noise and random noise to appear in the picture. This results in a maximum usable length for a given imager/cable/image processing electronics system.

Non-linear amplifiers located near the imager can be used to regenerate the original waveform for the reset-gate. However, this approach does not eliminate all of the noise generating mechanisms, and incurs size penalties at the imager head. This is unacceptable for medical and other devices which must be kept small.

SUMMARY OF THE INVENTION

Briefly stated, picture noise in a CCD video probe is reduced by producing a faithful reproduction of a reset-gate signal received from a timing generator to an imager by using a cable compensator network. The network contains an RLC circuit which corrects the relative amplitude and phase of the various frequency components of the reset-gate signal. The precise values of the components in the RLC circuit are selected according to the attenuation and distortion characteristics of the coaxial cable connecting the drive electronics assembly from the imager assembly.

According to an embodiment of the invention, a cable compensator network which receives a reset-gate signal from a timing generator and provides the reset-gate signal to an imager includes an RLC circuit which receives the reset-gate signal from a buffer circuit and corrects a relative amplitude and phase of a plurality of frequency components of the reset-gate signal; and a low-pass filter which receives the reset-gate signal from the RLC circuit and outputs the reset-gate signal to the imager.

According to an embodiment of the invention, a cable compensator network which receives a reset-gate signal from a timing generator and provides the reset-gate signal to an imager includes correcting means for performing phase and relative amplitude correction of the received reset-gate signal, wherein the correcting means including means for boosting high frequency components of the received reset-gate signal, means for attenuating low frequency components of the received reset-gate signal, and means for limiting attenuation of middle frequency components of the reset-gate signal.

According to an embodiment of the invention, a method for reducing picture noise in a CCD video probe includes the steps of (a) receiving a reset-gate signal from a timing generator; (b) performing phase and relative amplitude correction of the received reset-gate signal; wherein the step of performing includes the steps of boosting high frequency components of the received reset-gate signal, attenuating low frequency components of the received reset-gate signal,

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and limiting attenuation of middle frequency components of the reset-gate signal; and (c) providing the corrected reset-gate signal to an imager in the CCD video probe.

BRIEF DESCRIPTION OF THE DRAWINGS

FIG. 1 shows a top level schematic of the present invention.

FIG. 2 shows a detailed schematic of an embodiment of the present invention.

FIG. 3 shows a detailed schematic of an alternate embodiment of the present invention.

FIG. 4 shows a detailed schematic of an alternate embodiment of the present invention.

FIG. 5 shows an uncorrected reset-gate pulse measured at the head end as the pulse comes out of the cable.

FIG. 6 shows a corrected reset-gate pulse measured at the head end as the pulse comes out of the cable.

FIG. 7 shows a corrected reset-gate pulse measured at the head end as the pulse comes out of a shorter cable than that used for the measurement shown in FIG. 4.

DETAILED DESCRIPTION OF THE PREFERRED EMBODIMENT

Referring to FIG. 1, a drive electronics assembly 10 is connected to an imager assembly 30 via a coaxial cable 20. Imager assembly 30 includes a horizontal clocks processing section 32, a video processing section 34, and an imager section 36. Sections 32 and 34 can be contained within imager assembly 30 because they don't add significantly to the size of imager assembly 30. That is, they don't contain any inductors or large capacitors. In borescope, bronchoscope, and angioscope applications, drive electronics assembly 10 is not contained in or co-located with imager assembly 30 due to the need to keep imager assembly 30 as small as possible to facilitate proper movement within the pipes, tubes, bronchia, or veins/arteries that these instruments are designed for.

Drive electronics assembly 10 includes a waveform generator 12 which provides a reset-gate signal 14 and other clock signals 16 to imager assembly 30. Drive electronics assembly 10 further includes an image processing circuit 18 that processes the signals received from imager assembly 30. Reset-gate signal 14 passes through a cable compensator 40 which corrects the relative amplitude and phase of the various frequency components of the reset-gate signal.

Referring to FIG. 2, both the waveform generator 12 (FIG. 1) and the image processing circuit 18 (FIG. 1) are located in a separate assembly (not shown). In this embodiment, the reset-gate signal 14 passes through a buffer U3 and a DC blocking capacitor C4 before entering cable compensator 40 as signal RG2. In this case, the buffer U3 provides the necessary power to drive cable 20, but does not amplify reset-gate signal 14.

In essence, cable compensator 40 pre-distorts reset-gate signal 14 to compensate for the attenuation that occurs at different frequencies through cable 20. Cable compensator 40 includes capacitors C67, C54, C66; resistors R4, R44, R54; and inductors L54, L5. In this embodiment, R44, L5, and C66 form a low-pass filter end section that serves as part of the termination impedance for cable 20 as well as an EMI filter. R4, R44, and the output impedance of U3 provide low frequency termination impedance.

With a fundamental frequency of the reset-gate signal of 10–15 MHZ, this frequency range is considered to be "middle frequencies", with anything lower being a "low frequency" and anything higher being a "high frequency." The frequencies of particular concern for cable compensation are the fundamental frequency and the 2nd through 10th

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harmonics of the fundamental frequency. The primary components performing the relative phase and amplitude correction on reset gate signal **14** are series branch **C67** and **R44** and shunt branch **L54**, **R54**, and **C54**. **C67** and **L54** boost the amplitudes of the higher frequencies relative to the fundamental frequency. **R44** and **R54** limit the amount of boost from **C67** and **L54** respectively. **R54** and **C54** limit the attenuation that **L54** causes at lower frequencies. At middle and high frequencies, **C67**, **L54**, **L5**, and **C66** alter the matching impedance provided by **R44** and **R4**. In this implementation, tradeoffs are made between matching impedance and other parameters so that the impedance at these frequencies is lower than ideal.

Referring to FIG. 3, an alternate embodiment is shown in which shunt branch **L54**, **R54**, and **C54** are to the input side of the **C67**, **R4** combination.

Referring to FIG. 4, an alternate embodiment is shown in which buffer **U3** is after the "T"-section instead of before it. This embodiment permits obtaining an ideal matching impedance, but is more costly to implement.

Referring to FIG. 5, an uncorrected reset-gate pulse is shown measured at the head end (imager assembly end) of the borescope as the pulse comes out of cable **20**. The H-clock pulse is also shown. Cable **20** is 10 meters long. The waveform shows a long undesired asymptote after the falling edges.

Referring to FIG. 6, the corrected reset-gate pulse at the head end of the borescope after the pulse has been modified by cable compensator **40** is shown. Cable **20** is 10 meters long. The waveform falls promptly to the low level after the pulse.

Referring to FIG. 7, the corrected reset-gate pulse is shown when cable **20** is only 5 meters long. The waveform is nearly ideal except for a negative-going pulse that follows the main pulse. This is due to the non-ideal termination impedance of cable compensator **40**, but is of no harm to its operation.

While the present invention has been described with reference to a particular preferred embodiment and the accompanying drawings, it will be understood by those skilled in the art that the invention is not limited to the preferred embodiment and that various modifications and the like could be made thereto without departing from the scope of the invention as defined in the following claims.

What is claimed is:

1. A cable compensator network which receives a reset-gate signal from a timing generator and provides said reset-gate signal to an imager, comprising:

an RLC circuit which corrects a relative amplitude and phase of a plurality of frequency components of said reset-gate signal; and

a low-pass filter which receives said reset-gate signal from said RLC circuit and outputs said reset-gate signal to said imager.

2. A network according to claim 1, wherein said RLC circuit includes:

a first branch including a first capacitor in series with a first resistor and a first inductor;

a second branch including a second resistor in parallel with a second capacitor;

a third branch including a third resistor; and

said first, second, and third branches being connected as a "T", with said first branch positioned as a shunt to a reference potential.

3. A network according to claim 2, wherein said low-pass filter includes:

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a second inductor connected in series to said third resistor; a third capacitor connected between said second inductor and said reference potential; and

an output of said low-pass filter taken from between said second inductor and said third capacitor, said low-pass filter output being connected to said imager.

4. A network according to claim 3, further including a buffer and a fourth capacitor connected in series between said timing generator and said second branch.

5. A network according to claim 4, wherein said network is a component of one of a borescope, bronchoscope, and angioscope.

6. A network according to claim 2, wherein said third branch further includes a fourth capacitor in series with said third resistor.

7. A network according to claim 6, wherein said network is a component of one of a borescope, bronchoscope, and angioscope.

8. A network according to claim 2, wherein said third branch further includes a buffer in series with said third resistor.

9. A network according to claim 8, wherein said network is a component of one of a borescope, bronchoscope, and angioscope.

10. A network according to claim 1, wherein said network is a component of one of a borescope, bronchoscope, and angioscope.

11. A network according to claim 1, wherein said cable compensated by said network is at least 5 meters long.

12. A network according to claim 11, wherein said cable is 10 meters long.

13. A cable compensator network which receives a reset-gate signal from a timing generator and provides said reset-gate signal to an imager, comprising correcting means for performing phase and relative amplitude correction of said received reset-gate signal, wherein said correcting means boosts high frequency components of said received reset-gate signal relative to middle frequency components, limits attenuation of middle frequency components of said reset-gate signal, and passes low frequency components of said reset-gate signal.

14. A network according to claim 13, wherein said network is a component of one of a borescope, bronchoscope, and angioscope.

15. A network according to claim 13, further comprising means for filtering electromagnetic interference.

16. A network according to claim 15, wherein said network is a component of one of a borescope, bronchoscope, and angioscope.

17. A method for reducing picture noise in a CCD video probe that is a component of one of a borescope, bronchoscope, and angioscope, comprising the steps of:

receiving a reset-gate signal from a timing generator;

performing phase and relative amplitude correction of said received reset-gate signal; wherein the step of performing includes the steps of boosting high frequency components of said received reset-gate signal relative to middle frequency components, and limiting attenuation of middle frequency components of said reset-gate signal; and

providing said corrected reset-gate signal to an imager in said CCD video probe.

18. A method according to claim 17, further including the step of passing low frequency components of said reset-gate signal to said imager.

* * * * *

专利名称(译)	用于CCD视频探头的电缆补偿电路		
公开(公告)号	US6590470	公开(公告)日	2003-07-08
申请号	US09/592345	申请日	2000-06-13
[标]申请(专利权)人(译)	伟伦公司		
申请(专利权)人(译)	伟伦，INC.		
当前申请(专利权)人(译)	GE检测科技，LP		
[标]发明人	BURDICK KENNETH J		
发明人	BURDICK, KENNETH J.		
IPC分类号	H04N5/217 H04N5/232 H04N5/335 H04N5/225 A61B1/04 H04N5/357 H04N5/372 H04N5/376 H03H7/30 A61B17/36		
CPC分类号	H04N5/23203 H04N5/335 H04N5/357 H04N5/3765 H04N2005/2255		
审查员(译)	PASCAL，ROBERT		
助理审查员(译)	JONES，STEPHEN		
外部链接	Espacenet USPTO		

摘要(译)

通过使用电缆补偿器网络产生从定时发生器接收到的重置门信号到成像器的忠实再现，减少了CCD视频探头中的图像噪声。该网络包含RLC电路，该电路校正复位门信号的各种频率分量的相对幅度和相位。根据连接驱动电子组件与成像器组件的同轴电缆的衰减和失真特性来选择RLC电路中的元件的精确值。

