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(54) **HEART RATE ESTIMATION APPARATUS
WITH STATE SEQUENCE OPTIMIZATION**

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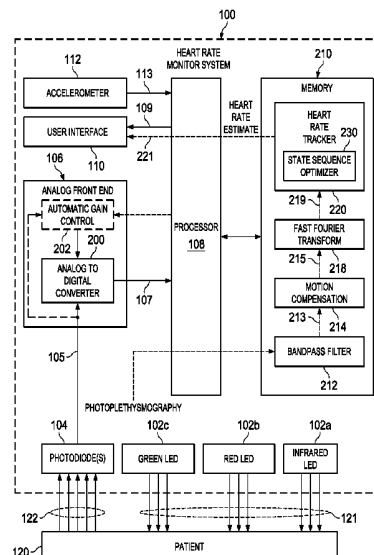
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(57) **ABSTRACT**

Disclosed examples include heart rate monitor systems and methods to estimate a patient heart rate or rate of another pulsed signal, in which rate hypotheses or states, are identified for a current time window according to digital sample values of the pulsed signal for a current sample time window, and a rate change value is computed for potential rate transitions between states of the current and previous time windows. Transition pair branch metric values are computed as a function of the rate change value and a frequency domain amplitude of the corresponding rate hypothesis for the current time window, and the pulsed signal rate estimate is determined according to a maximum path metric computed according to the branch metric value and a corresponding path metric value for the previous time window.

10 Claims, 9 Drawing Sheets



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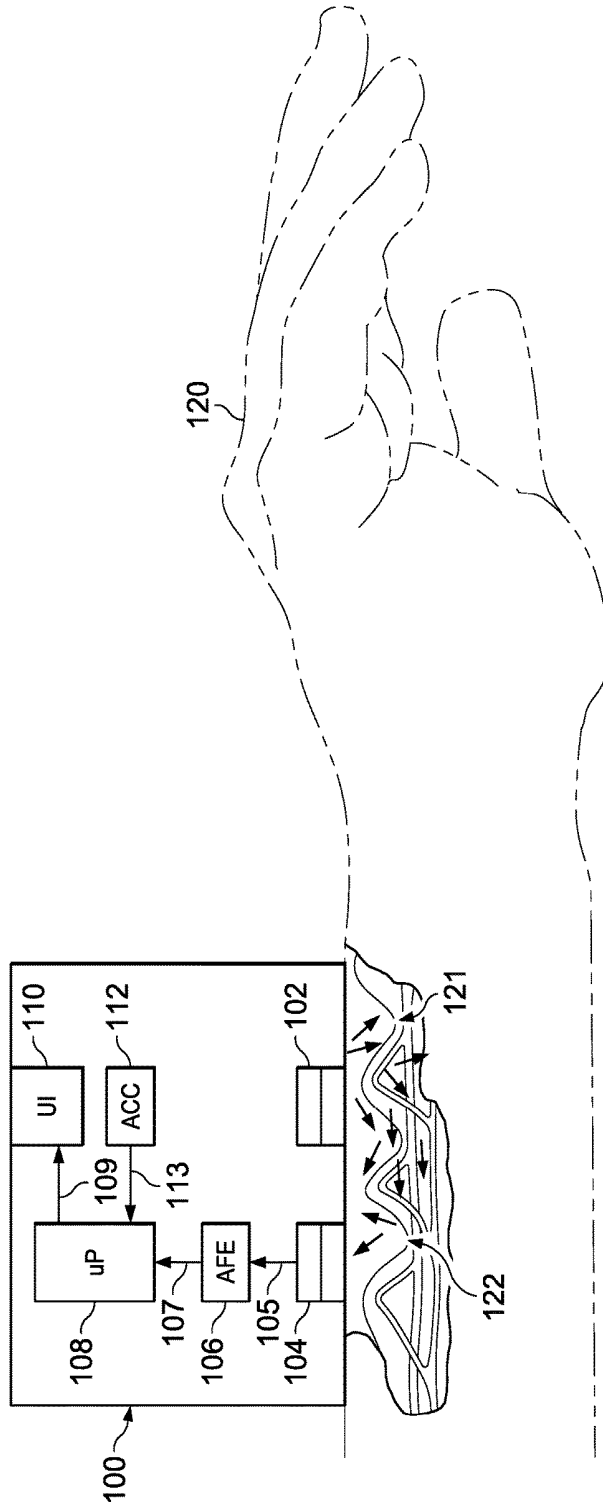
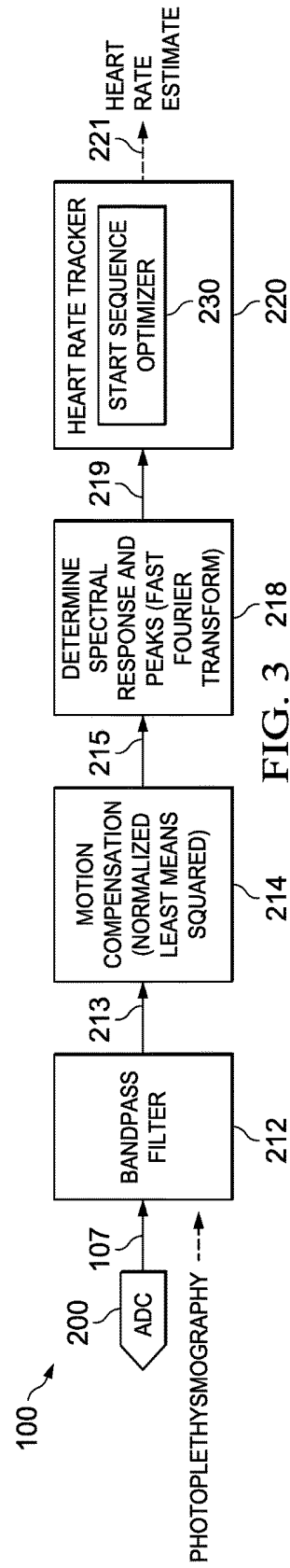


FIG. 1



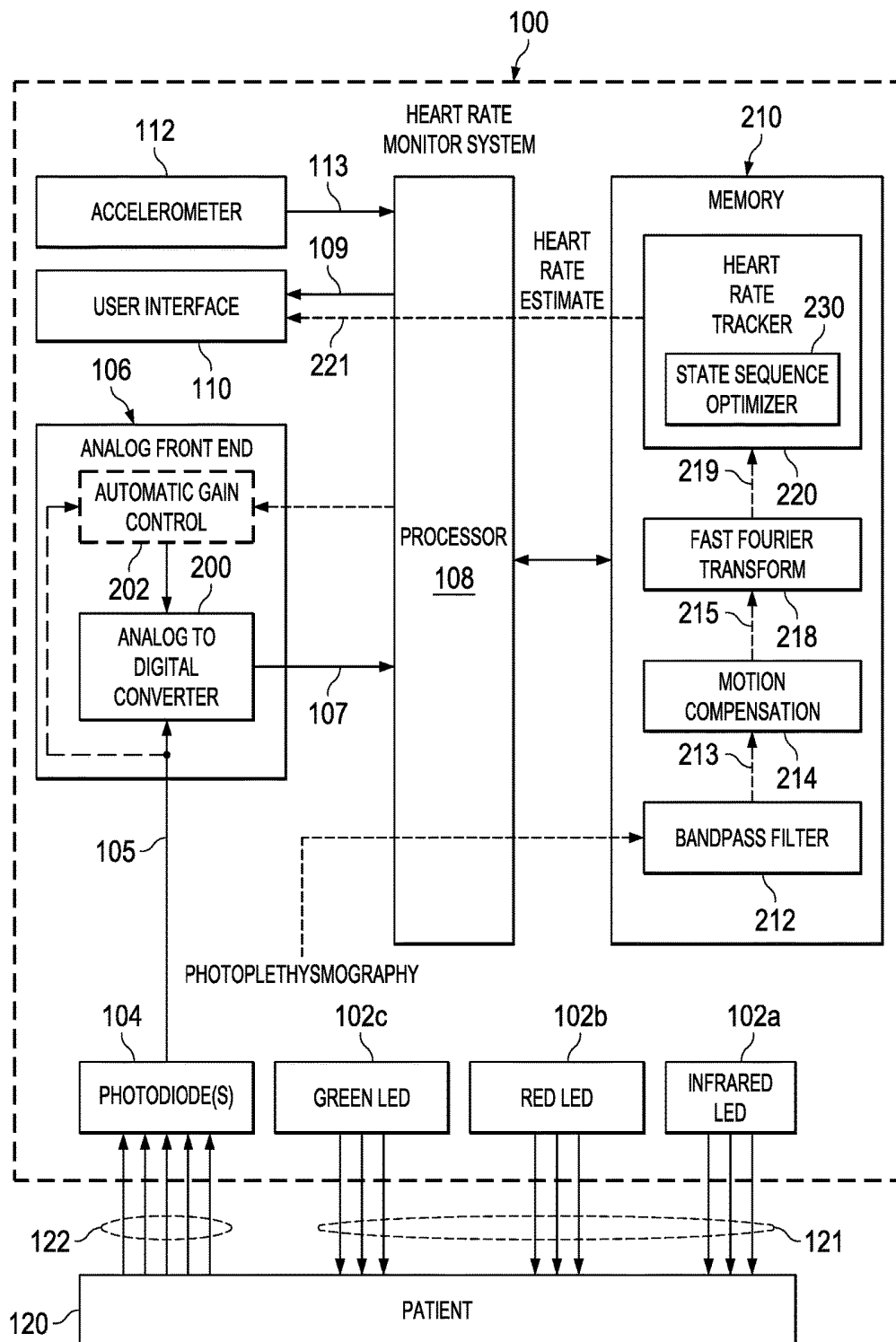
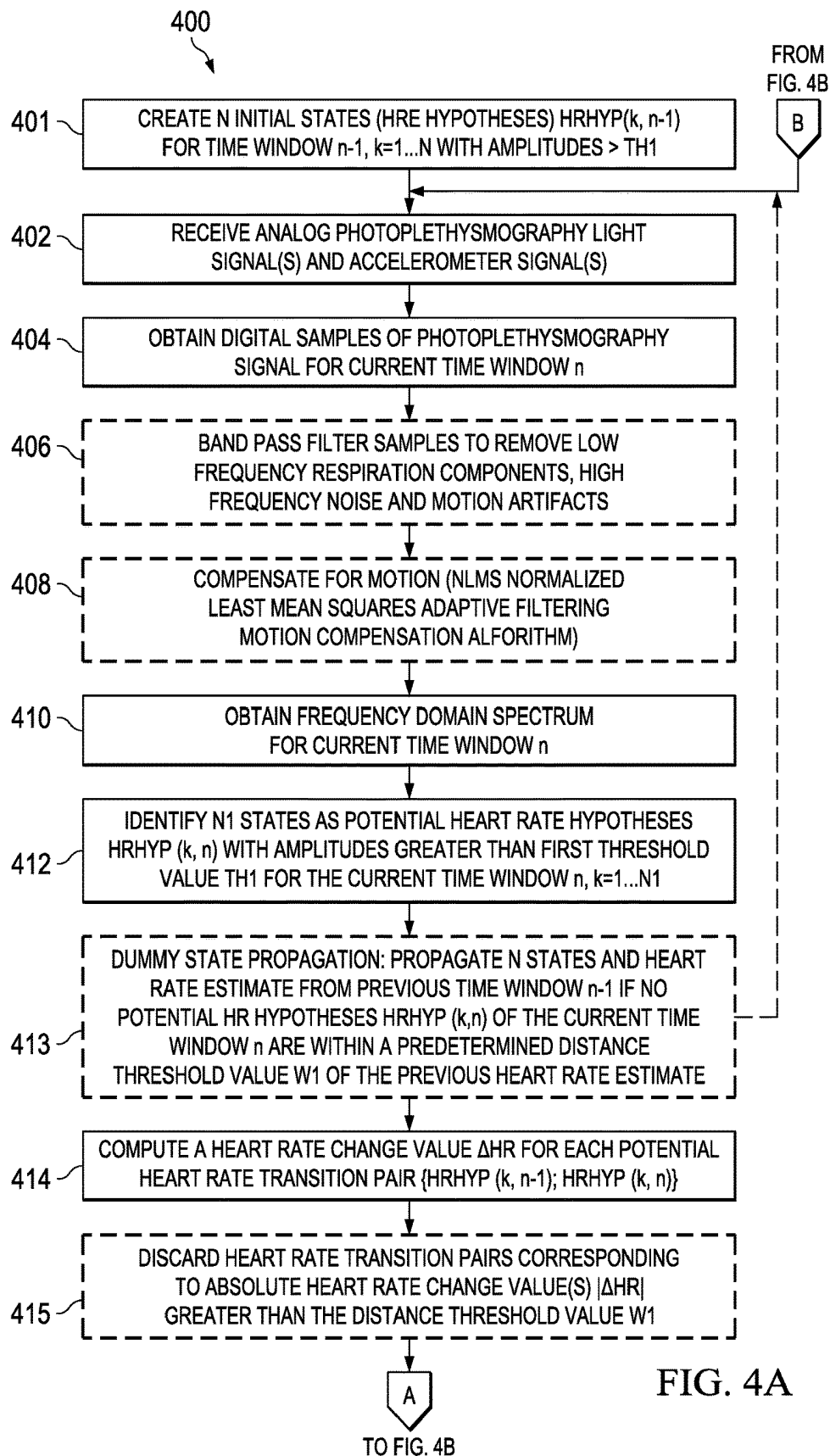


FIG. 2



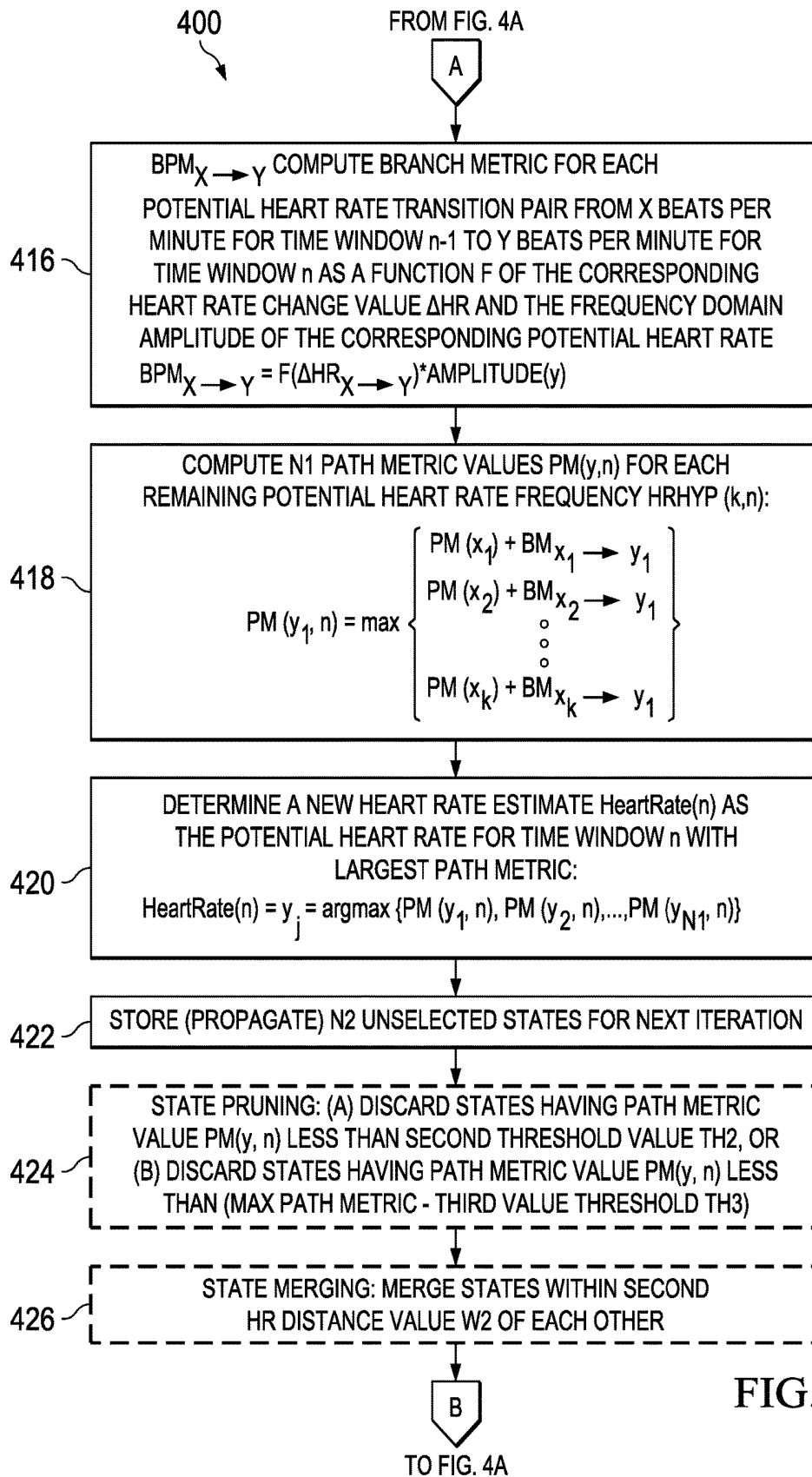
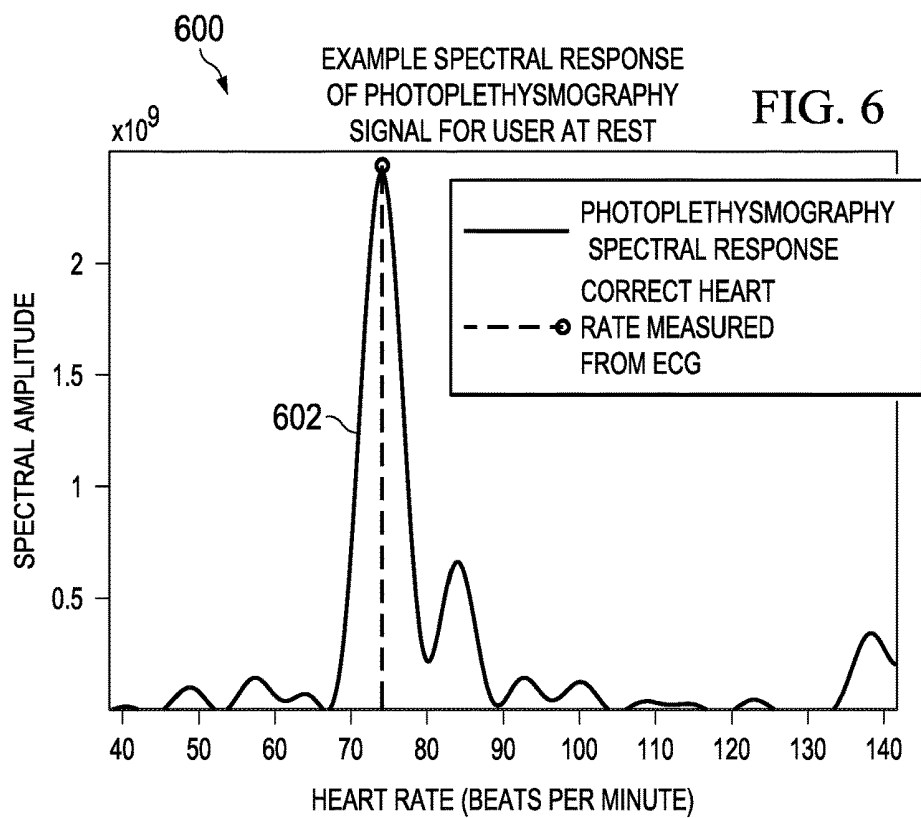
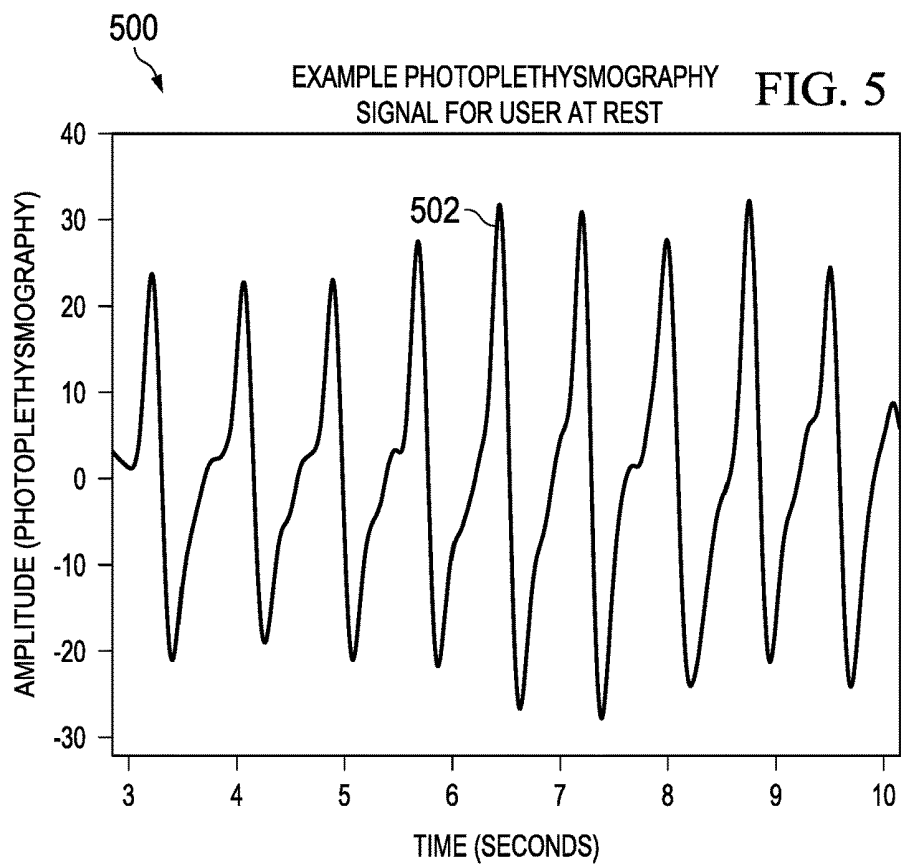
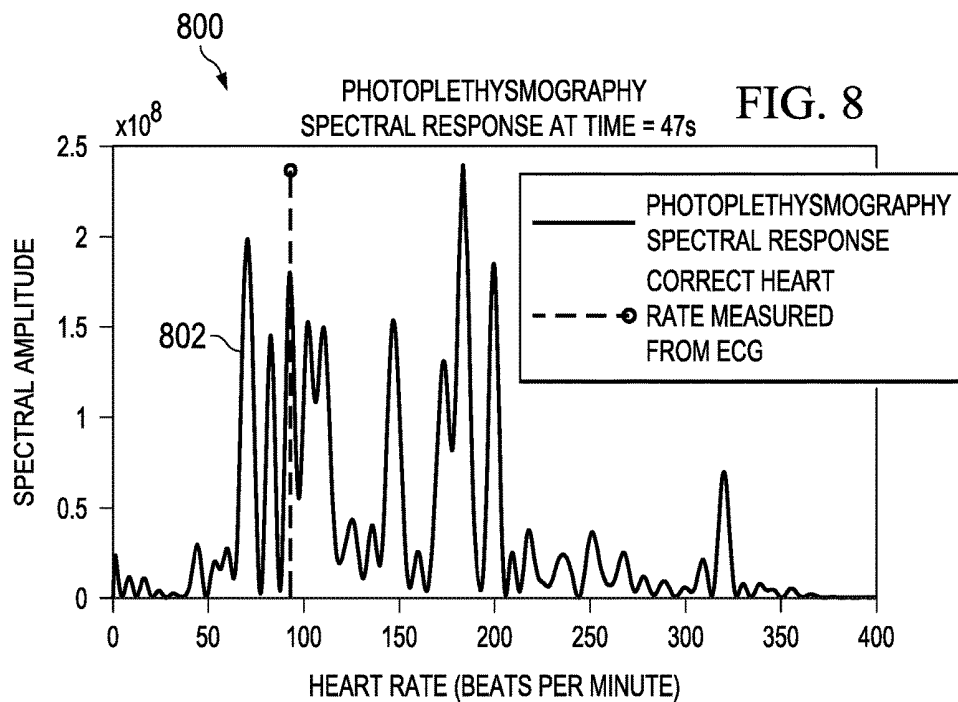
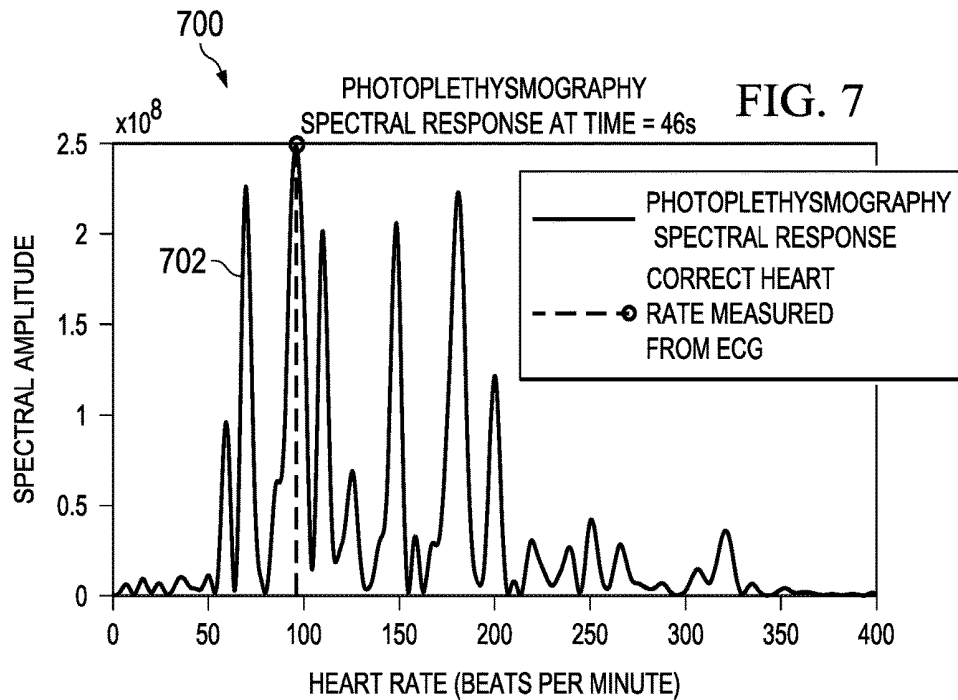


FIG. 4B





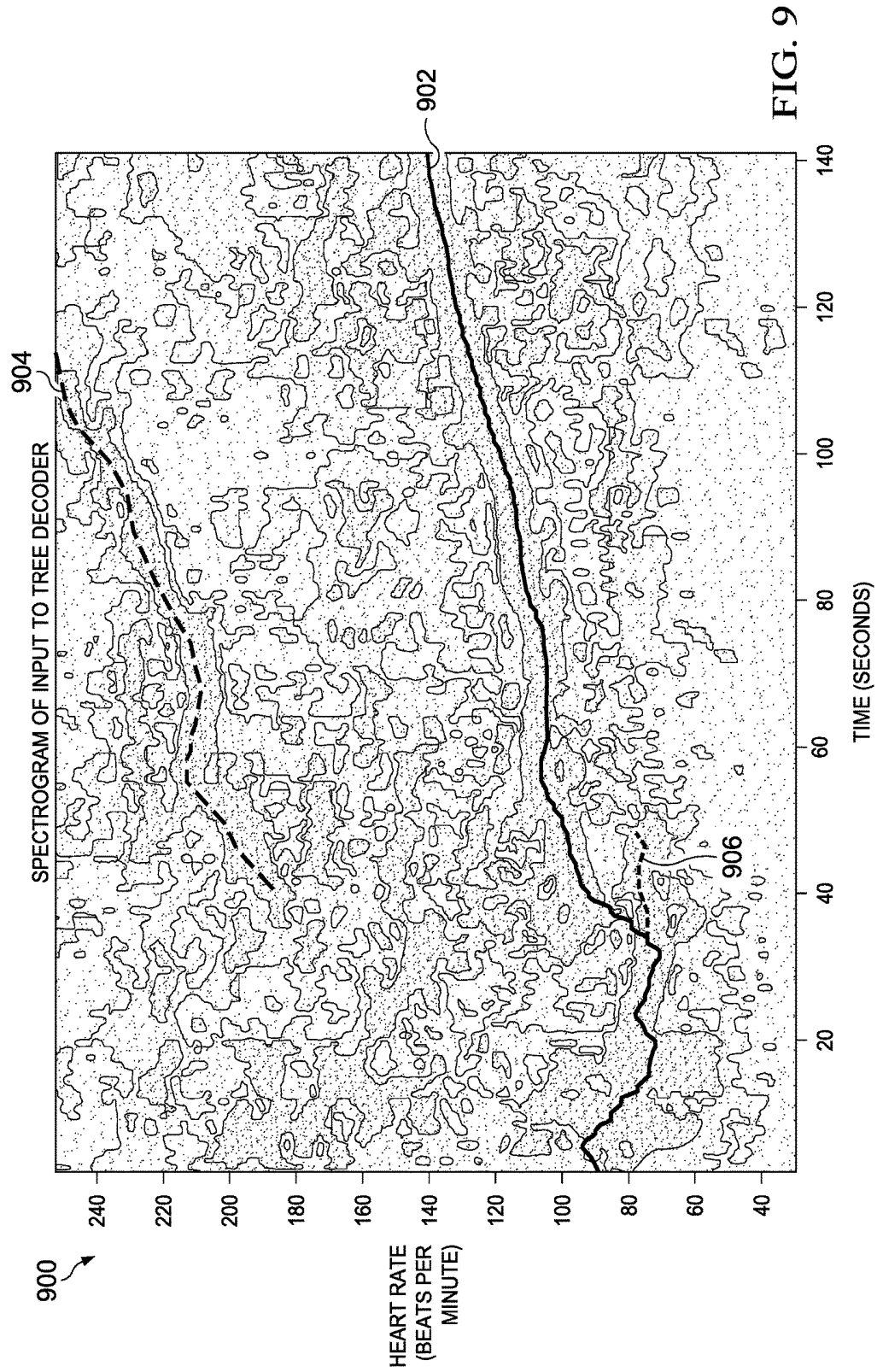
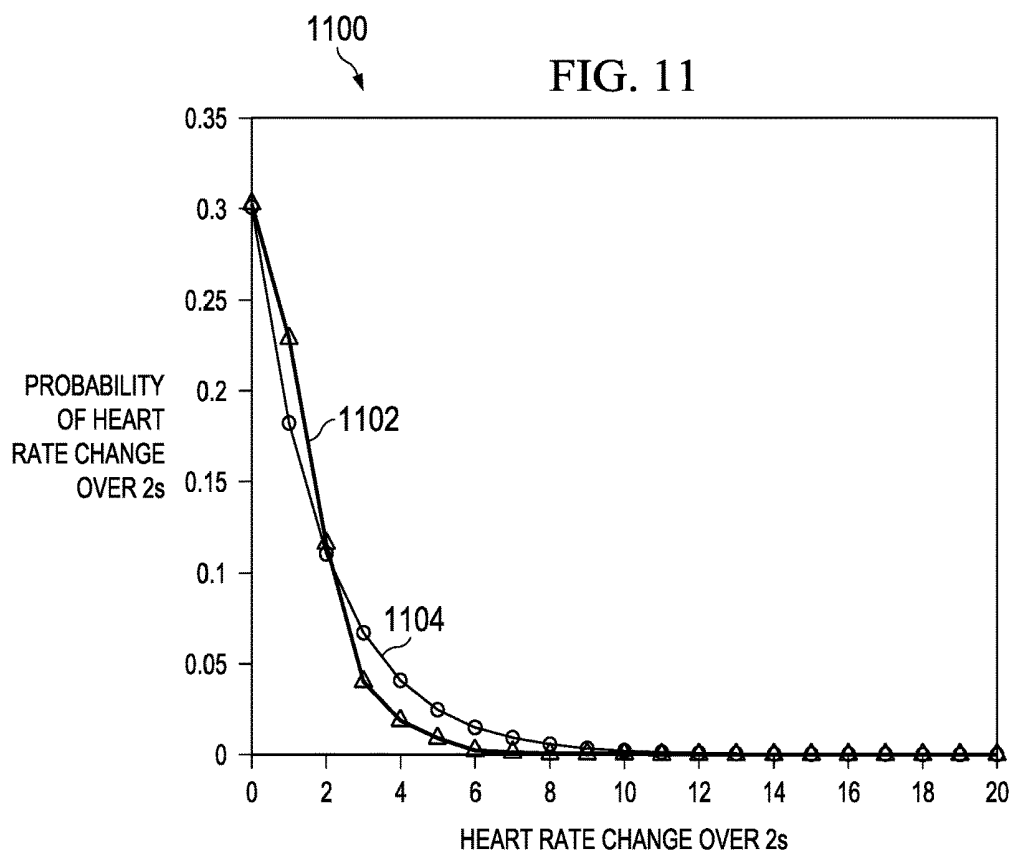
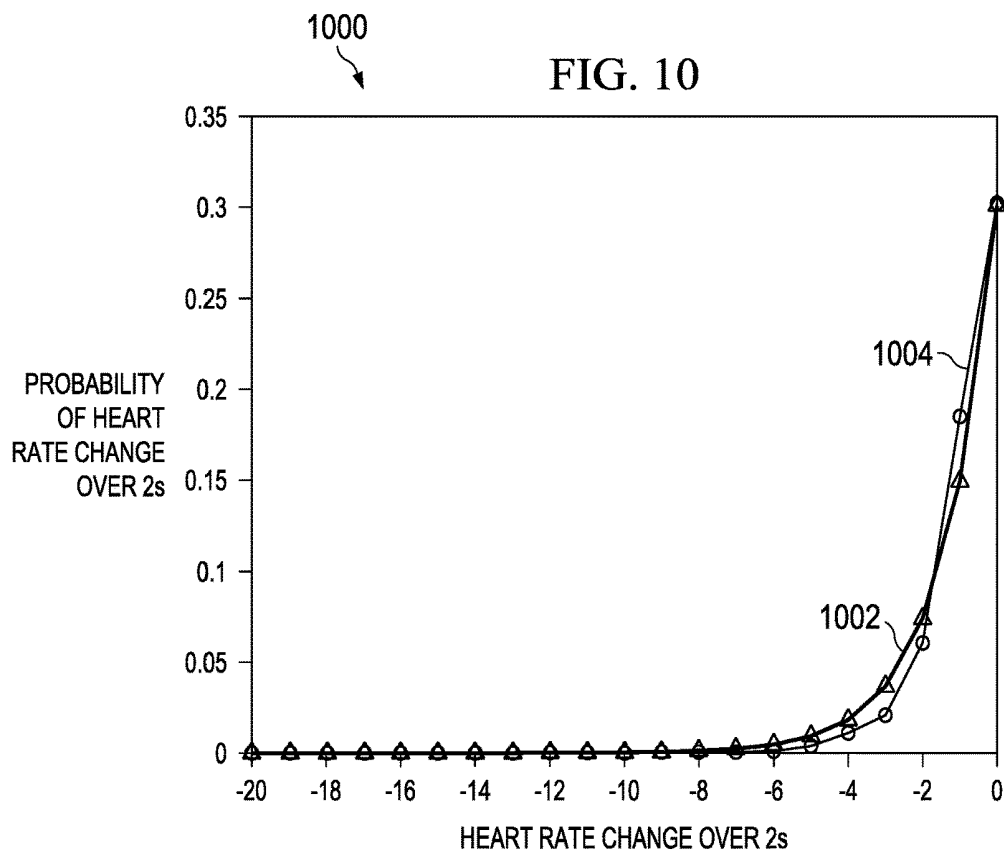
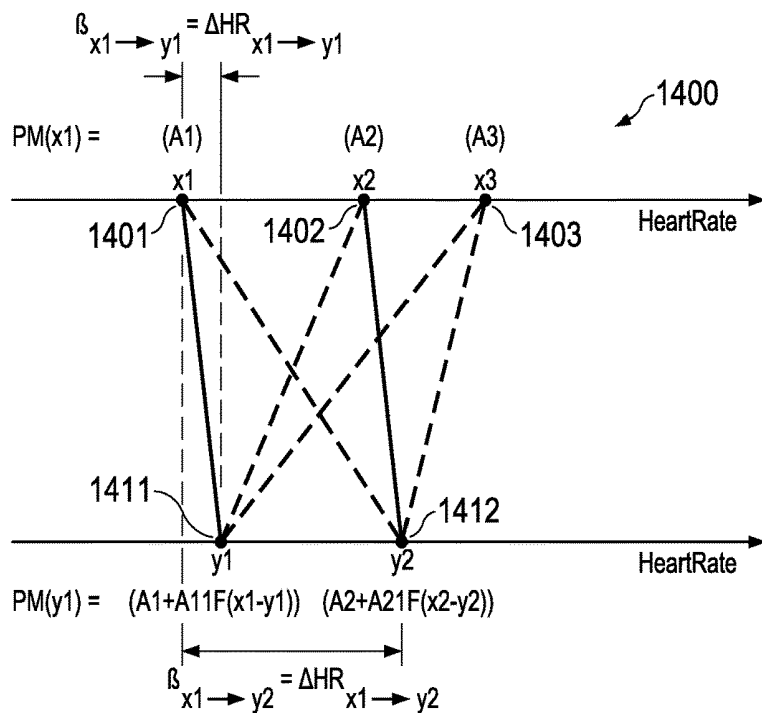
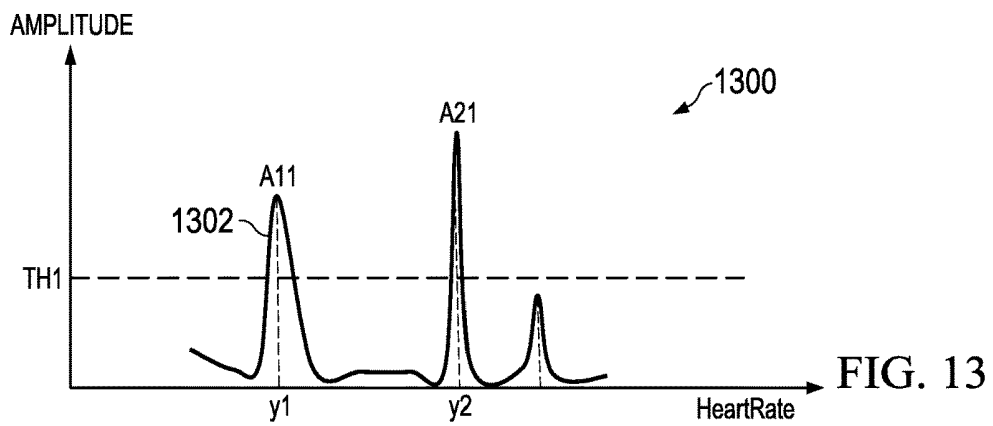
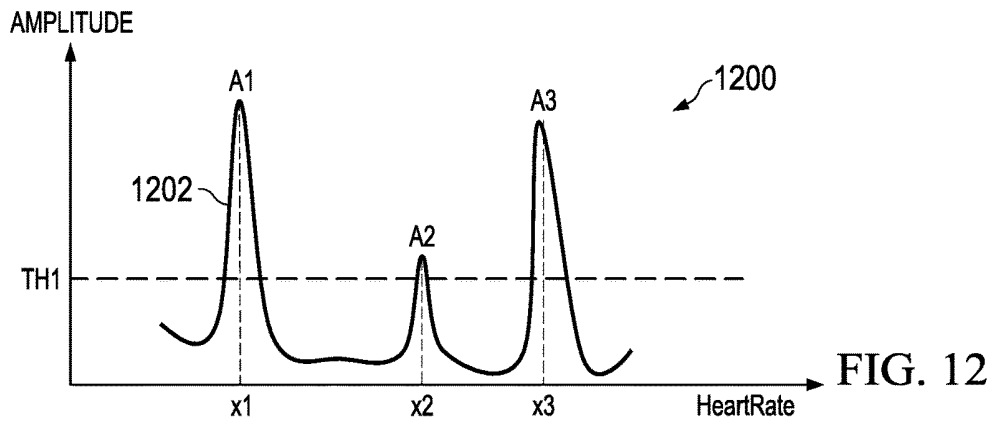


FIG. 9





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HEART RATE ESTIMATION APPARATUS WITH STATE SEQUENCE OPTIMIZATION

Under 35 U.S.C. § 119, this application claims priority to, and the benefit of, U.S. Provisional Patent Application Ser. No. 62/260,029 that was filed on Nov. 25, 2015 and is entitled “HEART RATE ESTIMATION WITH STATE SEQUENCE OPTIMIZATION”, the entirety of which is incorporated by reference herein.

BACKGROUND

A photoplethysmogram (“PPG”) is an optically obtained volumetric measurement of an organ (an optical plethysmogram). Photoplethysmography can be used in wearable activity monitors, medical equipment or other systems to optically detect blood volume changes in blood vessels to monitor blood flow, blood content, respiration rate and other circulatory conditions, where the intensity of back scattered light correlates to the amount of blood volume. PPG signals can be obtained in a number of different ways, including assessing absorption of light transmitted through, or reflected from, a patient’s skin. A light source at a particular wavelength (typically, red, infrared or green) directs light toward the patient’s skin. A photodiode or other optical sensor generates the PPG signal indicating the measured light transmission or reflection, and changes in the PPG signal can be used to detect the pulse rate of the patient’s heart. PPG based heart rate estimation during motion is difficult, as motion artifacts show up in the PPG signal. The motion artifacts are caused by hemodynamic effects, tissue deformation, and sensor movement relative to the skin. Motion compensation techniques have been proposed to remove the motion component in the PPG signal using information from an external sensor reference, such as an accelerometer. Some approaches use spectrum subtraction to first remove the spectrum of the acceleration data from that of the PPG signal prior to heart rate estimation. Another motion compensation approach uses compressed sensing techniques combined with signal decomposition for denoising and spectral tracking. The PPG signal fidelity can be further improved using normalized least mean squares (NLMS) and non-coherent combination in the frequency domain. A patient heart rate can be estimated from PPG signals using time-domain analysis, for example, band-pass filtering with a zero crossing detection, but this approach is largely unsuitable in the presence of motion artifacts. Frequency domain analysis generally involves choosing a highest peak in a frequency spectrum as the estimated patient heart rate. However, the PPG frequency spectrum typically includes multiple peaks within a range of realistic patient heart rates, and the dominant frequency domain peak does not always correspond to the actual patient heart rate, particularly in the presence of motion artifacts.

SUMMARY

Disclosed examples include heart rate monitor systems and methods to estimate a patient’s heart rate or a rate of another pulsed signal using state sequence tracking. Heart rate hypotheses, or states, are identified for a current time window according to digital sample values of the pulsed signal for a current sample time window, and a rate change value is computed for potential rate transitions between states of the current and previous time windows. Transition pair branch metric values are computed as a function of the rate change value and a frequency domain amplitude of the

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corresponding rate hypothesis for the current time window, and the pulsed signal rate estimate is determined according to a maximum path metric computed according to the branch metric value and a corresponding path metric value for the previous time window.

BRIEF DESCRIPTION OF THE DRAWINGS

FIG. 1 is a partial schematic diagram of a system to measure reflected light from a patient’s skin for estimating the patient heart rate.

FIG. 2 is a detailed schematic diagram of an example heart rate estimation system using state sequence optimization.

FIG. 3 is a schematic diagram of a heart rate estimation example for processing a single light signal in the system of FIGS. 1 and 2.

FIGS. 4A and 4B show a flow diagram of a method to estimate the heart rate of a patient.

FIG. 5 is a signal diagram of a time domain filtered PPG signal.

FIG. 6 is a signal diagram of a frequency domain filtered PPG signal corresponding to the time domain signal of FIG. 5 showing a single dominant peak corresponding to an actual patient heart rate.

FIG. 7 is a signal diagram of a frequency domain PPG signal spectrum at a first time when a dominant frequency domain peak corresponds to the correct patient heart rate.

FIG. 8 is a signal diagram of a frequency domain PPG signal spectrum at a second time when the dominant frequency domain peak does not correspond to the correct patient heart rate.

FIG. 9 is a short-term spectrogram showing frequency domain intensity for different heart rates over time, including a true patient heart rate tracking curve as well as alternate heart rate state tracks.

FIG. 10 is a graph of actual and modeled probability density function curves for decreasing or negative heart rate changes.

FIG. 11 is a graph of actual and modeled probability density function curves for increasing or positive heart rate changes.

FIG. 12 is a signal diagram of a frequency domain PPG signal spectrum at a first time including three frequency domain peaks above a threshold.

FIG. 13 is a signal diagram of a frequency domain PPG signal spectrum at a second time including two frequency domain peaks above the threshold.

FIG. 14 is a heart rate transition diagram showing possible transitions from the three states at the first time in FIG. 12 to the two states at the second time in FIG. 13.

DETAILED DESCRIPTION

In the drawings, like reference numerals refer to like elements throughout, and the various features are not necessarily drawn to scale. In the following discussion and in the claims, the terms “including”, “includes”, “having”, “has”, “with”, or variants thereof are intended to be inclusive in a manner similar to the term “comprising”, and thus should be interpreted to mean “including, but not limited to . . .”

Referring initially to FIGS. 1 and 2, FIG. 1 shows a wearable heart rate monitor system 100 to measure reflected light from the skin of a patient 120 to determine a heart rate estimate (HRE) value. FIG. 2 shows further details of the heart rate estimation system 100, including a processor

implemented heart rate tracker **220** with a state sequence optimizer (SSO) **230**. The system **100** estimates the patient heart rate by identifying heart rate hypotheses, or states, for a current time window according to digital sample values of a PPG signal or other pulsed signal for the current sample time window. The system **100** computes a rate change value for potential rate transitions between states of the current time window and the previous time window. The system **100** computes branch metric values for state transition pairs as a function of the rate change value and a frequency domain amplitude of the corresponding rate hypothesis for the current time window *n*, and determines the pulsed signal rate estimate according to a maximum path metric computed according to the branch metric value and a corresponding path metric value for the previous time window.

In one non-limiting pulsed signal rate estimation application, the system **100** is shown in FIG. 1 adjacent to a patient wrist for reflective PPG monitoring. In this example, the system **100** uses reflected light measurement techniques via one or more light sources (e.g., LED(s)) **102** and one or more optical sensors (e.g., photo diode(s)) **104**. The light source **102** and the optical sensor **104** are disposed along a common plane in this example. Coplanarity of the light source **102** and the sensor **104** is not required in all implementations. In other examples, transmission configurations are possible, with a light source and a light sensor disposed on opposite sides of a patient's finger or other portions of a patient's body to assess light transmission and/or absorption for determining patient heart rate. The light source **102** in FIG. 1 directs a first light signal **121** toward the skin of the patient **120**. The first light signal **121** in this example is reflected in the patient **120** with some measure of absorption, and a reflected second light signal **122** is generated. The optical sensor **104** detects the second light signal **122** from the patient **120**. In this example, the sensor **104** generates an analog PPG signal **105** that represents reflection of the first light signal **121** in the patient **120**. In other examples (not shown) where the sensor **104** and the source **102** are on opposite sides of the patient's finger, the sensor **104** provides a PPG signal **105** that represents transmission or absorption of light through the patient. In one example, a single light source **102** and a single optical sensor **104** are used. The example of FIG. 2 includes multiple LED light sources **102a-102c** that provide corresponding first light signals **121** at different wavelengths. The first light source **102a** in this example provides infrared (IR) light **121**, the second source **102b** provides red light **121**, and the third optical source **102c** provides green light **121**.

The system **100** further includes an analog front end (AFE) circuit **106**. One or more optical sensors **104** receive the second light signal **122**, and provide one or PPG signals **105** to the AFE circuit **106**. The AFE circuit **106** includes suitable analog signal conditioning circuitry (not shown) to receive and condition the PPG signal or signals **105**, as well as an analog to digital converter (ADC) **200**. The ADC **200** samples the PPG signal **105** and generates a plurality of digital PPG sample values **107** in a first time window for use in estimating the patient heart rate corresponding to the time window. The system **100** in one example operates in continuous fashion to continually monitor received PPG signals **105** and to estimate the patient heart rate based on an integer number of samples in individual ones of a series of multiple time windows to yield a heart rate estimate (HRE) value for each corresponding time window.

The ADC **200** operates at a fixed or adjustable sample rate, and the time windows in certain examples include a sufficient number of samples to characterize the PPG signal

105 over multiple heartbeats of the patient **120**. For example, the length of the time windows *W* in one example is on the order of 8-10 seconds. The time windows in one example are of equal length, but other embodiments can have time windows of different lengths. The AFE circuit **106** in FIG. 2 also includes an analog automatic gain control (AGC) circuit **202** to adjust the gain of the signal **105** converted by the ADC circuit **200**. In certain examples, the AGC circuit **202** receives the signal or signals **105** and dynamically adjusts an input buffer amplifier gain to provide a signal for conversion by the ADC circuit **200**. In some examples, the analog AGC circuit **202** may also receive an external control signal or value from a processor **108** to set the gain for conversion of the PPG signal or signals **105** based on digital automatic gain control (DAGC) operation in the digital domain on previously received samples. In addition, the system **100** may include an accelerometer **112** providing a signal **113** to the processor **108**.

The ADC **200** provides digital sample values **107** of the PPG signal to the processor **108** and an associated electronic memory **210** for digital processing to estimate the patient heart rate. The processor circuit **108** can be any suitable digital logic circuit, programmable or pre-programmed, such as an ASIC, microprocessor, microcontroller, FPGA, etc. that operates to execute program instructions stored in the electronic memory **210** to implement the features and functions described herein as well as other associated tasks to implement a monitor system **100**. In certain examples, moreover, the memory circuit **210** can be included within the processor circuit **108**. The processor **108** in the examples of FIGS. 1 and 2 includes an interface connection **109** to a user interface (UI) **110**, such as a display (not shown). In certain examples, the memory **103** constitutes a non-transitory computer-readable storage medium that stores computer-executable instructions that, when executed by the processor **108**, perform the various features and functions detailed herein.

In the example of FIG. 2, the processor **108** implements instructions corresponding to various functions or components **212-220** in the memory **210**, including a filter component (bandpass filter or BPF) **212**, a motion compensation component **214**, a Fast Fourier Transform (FFT) component **218** and a heart rate tracker (HRT) component **220** with a state sequence optimizer **230**. In operation, the processor **108** implements the functional components **212-230** to operate on the received PPG samples **107** to provide a heart rate estimate value (HRE) **221**, which can then be provided to the user interface (UI) **110**, such as a display to render the HRE value **221** to a patient **120**.

The processor **108** implements the filter component **212** to filter **706** the digital PPG sample values **107** to generate a plurality of filtered values **213** corresponding to the first time window *W1*. The processor **108** implements low pass filtering by the filter component **212** to remove certain motion artifacts or other low frequency components not related to the patient heart rate. In the illustrated example, the processor **108** executes the instructions of the component **212** to implement bandpass filtering to also remove high-frequency noise components. As previously discussed, the methods and systems of the present disclosure can be used for estimating a rate of any pulsed signal. Although described herein in the context of a pulsed PPG signal **105** heart rate estimation, the various concepts can be employed with any pulsed signal. The processor **108** in the illustrated example performs motion compensation processing on the filtered values **213** to generate a plurality of motion compensated values **215** corresponding to the first time window *n*. In one

example, the motion compensation component **214** implements any suitable motion compensation algorithms or processing, including spectrum subtraction, compressed sensing with signal decomposition, and/or normalized least means squared or NLMS processing.

Following motion compensation, the processor **108** provides implements the FFT component **218** and the HRT component **220** in order to determine the heart rate estimate HRE value **221** representing the heart rate of the patient **120** for the current time window *n* according to the frequency content of the motion compensated values **215**. In one example, the processor computes the frequency content of the motion compensated values **215** using a Fast Fourier Transform FFT algorithm via the instructions of the component **218** in order to generate a frequency spectrum including frequency component values **219**. The processor **108** implements the heart rate tracking component **220** in this example to determine the HRE value **221** according to discernible peaks in the frequency spectrum data **219** via the state sequence optimizer component **230**.

FIG. 3 illustrates the digital domain processing for a single PPG light signal **105** in the system **100** of FIGS. 1 and 2 as described above. In other examples, multiple light signals **105** can be processed using the components **212-220**, for example, based on first light signals **121** generated by light sources **102** of different wavelengths (e.g., infrared or IR, red, green, etc.). The system **100** in this example may include a corresponding set of multiple optical sensors (e.g., photodiodes) **104**, each receiving a corresponding light signal **122** and generating a corresponding second analog PPG signal **105** representing transmission or reflection in the patient **120**. For example, the first optical sensor **104** detects the second light signal **122** representing transmission or reflection of the first light signal **121** from the first light source **102a**, and the processor **108** operates on sampled values in a first signal processing chain associated with the first wavelength. For an additional light source **102b** and an associated second optical sensor **104**, the analog PPG signal **105** corresponds to a detected fourth light signal **122** representing transmission or reflection of a third light signal from the second source **102b**, and the processor **108** forms filtering and motion compensation (and possibly gain adjustment) in a second digital signal processing chain.

FIGS. 4A and 4B provide a flow diagram of a method **400** to estimate a rate of a pulsed signal. In the general case, the method **400** can be implemented in any suitable processing system. In one example, the processor **108** of the heart rate monitor system **100** is programmed by instructions in the memory **210** in order to implement the method **400** to estimate a patient heart rate based on a pulsed PPG signal **105**, and the method **400** as described below in the context of this example application. In other implementations, the process **400** can be implemented in a general purpose computer (not shown), in a medical equipment processor, or other processing circuit or system. The processor **108** of the heart rate monitor system **100** generally implements the method **400** in continuous fashion, providing a heart rate estimate at each of a series of sample time windows.

As shown in FIG. 4A, the processor **108** begins executing the state sequence optimizer (SSO) component **230** (e.g., at power up) to characterize an initial frequency domain spectrum of the PPG signal input by creating an integer number *N* initial states corresponding to a time window *n-1*. In this initial or training phase, the SSO **230** may be initialized to a maximum of *N* heart rate hypotheses or states, that have amplitudes in the frequency domain greater than a first threshold value for the time window *n-1*. In some embodi-

ments, the initial heart rate hypotheses may correspond to the largest *N* distinct spectral peaks whose amplitude is above a first threshold TH1A. For each heart rate hypothesis, a corresponding path metric is assigned and the metric may simply be the spectral amplitude. In one example, the processor **108** sets the initial path metric to the spectral amplitude normalized by a scaling factor. Since the heart rate typically includes second and third harmonics, in an alternative embodiment an initial search window may first be defined e.g., [30 BPM-250 BPM], and the initial *N* heart rate hypotheses correspond to the *N* largest distinct peaks in the search window that have second and third harmonics above a threshold TH1B. Alternatively, the *N* hypotheses may correspond to the *N* heart rates hypotheses in the initial search window with the largest metric, where the metric is the sum of fundamental spectral peak, second harmonic peak and third harmonic peak.

At **402** in FIG. 4A, the iterative heart rate estimation processing starts with the analog front end circuit **106** receiving one or more analog PPG light signals. In certain examples, accelerometer signals are also received at **402**. At **404**, digital samples are obtained of the analog PPG signal over a corresponding current time window *n*, for example, using an ADC circuit **200** as described above. At **406**, the digital PPG samples of the current time window *n* are filtered, for example, by execution of digital filtering program instructions **212** by the processor **108** described above. At **408**, the filtered samples are motion compensated. In one example, the motion compensation at **408** is implemented using NLMS adaptive filtering or other suitable algorithm by processor execution of the compensation component instructions **214** above to provide motion compensated values **215**. As discussed above, the method **400** can be used for other pulsed signals, and the filtering and motion compensation steps at **406** and **408** can optionally be omitted.

Referring briefly to FIGS. 5-8, a signal diagram **500** and FIG. 5 shows a pulsed or oscillating time domain filtered PPG signal **502**, and the diagram **600** and FIG. 6 shows a corresponding frequency domain filtered PPG signal **602** with a single dominant peak (e.g., at approximately 73 bpm) corresponding to an actual patient heart rate. PPG based heart rate estimation during motion is sometimes difficult due to motion artifacts in the PPG signal. This problem can be addressed, to a certain extent, using one or more motion compensation algorithms (e.g., at **408** in FIG. 4A) and/or use of a heuristic based tracker to track the spectral peak. However, heuristic based trackers require multiple FFTs, and thus are undesirable for low complexity implementations. In addition, heuristics for tracking the spectral peak do not provide a structured approach for candidate tracking. The state sequence optimizer **230** in the system **100** advantageously provides heart rate tracking that makes prior use of how the heart rate evolves over time, and inherently tracks potential solutions by associating a metric with the heart rate estimate that depends on the current time window *n* as well as history from one or more prior time windows *n-1*. In this regard, and a patient is at rest and there are no motion artifacts, the heart rate can be easily inferred from the PPG signal through either counting the zero-crossings or keeping track of the peaks and valleys. Thus, at rest, the dominant spectral component also corresponds to the heart rate as shown in FIG. 6. However, the PPG signal typically includes motion artifacts when the patient is not at rest. For example, if a watch containing the optical heart rate system **100** is not tied tightly to the patient's wrist, the watch moves when the user is in motion. In this case, the motion artifact effect is multiplicative. The additive components result in

new peaks in the spectral signature while the multiplicative effect results in a spectral spreading (multiplication in time is convolution in frequency) of the heart rate signature i.e., presence of new peaks.

Referring also to FIGS. 7 and 8, absent perfect motion compensation and a perfect deconvolution of the multiplicative artifacts, it is difficult to minimize the motion induced spectral peaks. Importantly, the maximum peak may not always correspond to the correct heart rate even after motion compensation processing. In addition, as the patient motion changes over time, this inaccuracy in peak tracking also changes over time. A signal diagram 700 and FIG. 7 illustrates a frequency domain PPG signal spectrum 702 at a first time when a dominant frequency domain peak corresponds to the correct patient heart rate. At a subsequent sample time window, a signal diagram 800 in FIG. 8 shows the frequency domain PPG signal 802 when the dominant frequency domain peak does not correspond to the correct patient heart rate. As seen in FIG. 8, at any given time window, there may also be several spectral peaks that have similar amplitude to the spectral peak that corresponds to the correct heart rate. However, as the motion component in the PPG signal is typically a non-stationary process, there is a higher likelihood of the dominant spectral peak or a set of the dominant spectral peaks to correspond to the correct heart rate. Furthermore, the heart rate change from one time instant to the next cannot arbitrarily jump by a large amount i.e., it follows a certain distribution.

Returning to FIG. 4A, the heart rate monitor system processor 108 implements the state sequence optimization component 230 in which each candidate heart rate is considered as a possible state and is assigned a probability or a path metric. The path metric preferably captures the history of the previous heart rate estimate(s), and captures the rate of change of the heart rate from the previous heart rate estimate. Processor execution of the SSO component 230 in the illustrated example uses the spectral response of the motion compensated PPG signal as an input. In general, the processor execution of the SSO component 230 creates the initial heart rate hypothesis based on the spectral amplitudes and assigns a metric to each hypothesis, and at each time step or sample time window, computes branch metrics (416 in FIG. 4B) from the heart rate hypothesis for time window n-1 to those for time window n. The process 400 also includes computing or updating a path metric at 418 for the heart rate hypothesis, and determines a new heart rate estimate for the current time window n at 420 representing the patient heart rate as the heart rate hypothesis having a maximum candidate path metric value. The SSO component 230 in certain examples also selectively provides dummy state propagation at 413 to carry a previous set of states forward where a subset of the most probable heart rate hypothesis for time window n-1 have no valid heart state to jump to in time window n, and may also optionally implement state pruning and state merging at 424 and 426 in FIG. 4B as discussed further below.

Referring also to FIGS. 9-14, operation of the state sequence optimizer component 230 is illustrated and described with respect to a simplified example using PPG signals for heart rate estimation. FIG. 9 shows a short-term spectrogram diagram 900 that includes frequency domain intensity for different heart rates (in beats per minute or bpm) over time, including a true patient heart rate tracking curve 902 as well as alternate heart rate state tracks 904 and 906. FIG. 10 shows a graph 1000 of actual and modeled probability density function curves 1002 and 1004, respectively for decreasing or negative heart rate changes. A graph

1100 in FIG. 11 shows curves 1102 and 1104 respectively representing actual and modeled probability density function curves for increasing or positive heart rate changes. FIGS. 10 and 11 illustrate monotonically decreasing probability density functions (PDFs) representing the decreasing probability that a patient heart rate will change by large amount within a given timeframe (e.g., two seconds in the illustrated examples). FIG. 12 illustrates a frequency domain PPG signal spectrum 1200 as a curve 1202 at a first time including three frequency domain peaks of amplitude A1, A2 and A3 at corresponding heart rates or frequencies x1, x2 and x3 above a first threshold value TH1. In one example, the spectrum 1200 of FIG. 12 represents PPG sample data from a previous sample time window n-1. FIG. 13 illustrates a subsequent frequency domain PPG signal spectrum as a curve 1302 at a current time window n including two frequency domain peaks at heart rates y1 and y2 having corresponding amplitudes A11 and A21 above the threshold TH1. The state sequence optimizer component 230 and the process 400 are described hereinafter with respect to the examples of FIGS. 12 and 13 for updating state transition path metrics to estimate a current patient heart rate, where FIG. 14 provides a heart rate transition diagram 1400 showing possible transitions from the three states 1401, 1402 and 1403 at the time window n-1 in FIG. 12 to the two states 1411 and 1412 at the current time window n in FIG. 13.

The state sequence optimization processing begins at 410 in FIG. 4A. At 410, the processor 108 obtains a frequency domain spectrum for the current time window n. In one example, the processor computes the frequency content of the motion compensated values 215 using a Fast Fourier Transform FFT algorithm at 410 to generate 410 a frequency domain spectrum for the current time window n according to the digital PPG sample values 107. At 412, the processor 108 identifies a plurality N1 of states for the current time window n as heart rate hypotheses HRHYP(k, n) having amplitudes in the frequency domain spectrum greater than a first threshold value TH1 for the current time window n. A kth two-tuple (HRHYP(k, n-1), PM(k, n-1))k denote the kth heart rate hypothesis (HRHYP(k, n-1)) at time window n-1 with a path metric equal to PM(k, n-1). At the current time window n, the spectral response of the motion compensated PPG signal is determined at 410.

At 412, the set of N1 heart rate hypotheses are identified from the spectral response whose amplitudes are largest and are distinct peaks i.e., [(HR(1, n), Amp(1, n)), (HR(12, n), Amp(2, n)), . . . , (HR(N1, n), Amp(N1, n))]. In one example, the spectral input is computed by an FFT on the motion compensated PPG signal, and the distinct peaks of amplitudes A11 and A12 in FIG. 13 are found by looking for all local maxima that are above the threshold TH1. The spectral input may be normalized in various ways prior to comparing against a threshold. Several non-limiting examples include normalizing with respect to the maximum peak, normalizing with respect to the total power in the input signal, and normalizing with respect to a predefined constant. As noted at 412 in FIG. 4A, the SSO component 230 in certain examples provides an efficient implementation after an initial set of heart rate hypothesis are identified. For each heart rate hypothesis or state, the search space is within a distance threshold value +/-W1 heart beats from the previous HRE estimate value. In this example, the processor 108 only computes only the spectral response around those frequencies. In one embodiment, the processor 108 imple-

ments this approach by the Goertzel algorithm. In another embodiment, this computation may be done by using the DFT itself.

At 413, the processor 108 optionally implements dummy state propagation, for example, for iterations of the process 400 in which no spectral peak is identified for time window n which is within the distance threshold W1 of the heart rate estimate for time window n-1. In this case, the processor 108 determines the HRE value 221 representing the heart rate of the patient 120 for the current time window n as the previous heart rate estimate HRE for the previous time window n-1. In addition, the processor 108 carries at least some of the heart rate hypotheses HRHYP (k, n-1) from the previous time window n-1 forward to the current time window n. In this regard, the number of hypotheses N from the previous time window n-1 may, but need not be, the same as the number of hypotheses N1 determined for the current time window n. In certain examples, where none of the spectral peaks are within the distance threshold W1, the heart rate estimate value 221 for time window n-1 is carried forward along with its path metric to time window n, and the top N2 candidates for time window n-1 with the largest path metric are carried forward.

At 414, the processor 108 computes a heart rate change value ΔHR for each of a plurality of potential heart rate transition pairs. As shown in FIG. 14, for example, arrowed lines represent individual heart rate transition pairs corresponding to a heart rate transition to one of the heart rate hypotheses HRHYP (k, n) for the current time window n from one of a plurality of heart rate hypotheses HRHYP (k, n-1) for a previous time window n-1. FIG. 14 illustrates two example heart rate change values $\Delta HR_{x1 \rightarrow y1} = \beta_{x1 \rightarrow y1}$ and $\Delta HR_{x1 \rightarrow y2} = \beta_{x1 \rightarrow y2}$ for potential transitions from the point x1 at the previous time window n-1 to the points y1 and y2 at the current time window n. In one implementation, the processor 108 selectively discards heart rate transition pairs at 415 which correspond to heart rate change values ΔHR that are greater than the predetermined distance threshold value W1. In this example, for each heart rate hypothesis at the current time window n, the processor 108 computes the heart rate change values ΔHR for all the heart rate hypotheses for time window n-1, and only retains the heart rate change values ΔHR that are less than a distance threshold W1 are considered as valid transitions according to the following formulas:

$$\begin{aligned} \Delta HR_{1 \rightarrow 1} &= HRHYP(1, n-1) - HR(1, n); \\ 1 \rightarrow 1 \text{ transition valid if } \text{abs}(\Delta HR_{1 \rightarrow 1}) &< W1 \\ \Delta HR_{k \rightarrow 1} &= HRHYP(k, n-1) - HR(1, n); \\ k \rightarrow 1 \text{ transition valid if } \text{abs}(\Delta HR_{k \rightarrow 1}) &< W1 \\ &\vdots \\ \Delta HR_{k \rightarrow N1} &= HRHYP(k, n-1) - HR(N1, n); \\ k \rightarrow N1 \text{ transition valid if } \text{abs}(\Delta HR_{k \rightarrow N1}) &< W1 \end{aligned}$$

As an example, if a heart rate hypothesis for time window n is 70 BPM, and the distance threshold value W1 is set to 20 BPM, the processor 108 only considers the heart rate hypothesis at the previous time window n-1 which are between 50 BPM and 90 BPM as valid transitions to the heart rate hypothesis of 70 BPM at the current time window n.

Returning to 416 in FIG. 4B, the processor 108 computes branch metric values $BM_{x \rightarrow y}$ for each of the remaining plurality of potential heart rate transition pairs as a function F of the corresponding heart rate change value ΔHR and the frequency domain amplitude AMPLITUDE(y) of the corresponding heart rate hypotheses HRHYP (k, n) for the current time window n. In one example, the processor 108 computes the branch metric $BM_{x \rightarrow y}$ at 416 for each of the potential heart rate transition pairs as a product of the function F of the corresponding heart rate change value ΔHR and the frequency domain amplitude AMPLITUDE(y) of the corresponding heart rate hypotheses HRHYP (k, n) for the current time window n. In one implementation, for potential heart rate transition $\Delta HR_{x \rightarrow y}$, a heart rate hypothesis x at the previous time window n-1 is represented as HRHYP (x, n-1) to a heart rate hypothesis y for time window n HRHYP (y, n-1), the branch metric $BM_{x \rightarrow y}$ is defined as $F(\Delta HR_{x \rightarrow y}) * AMP(y)$. In this example, the weighting function F(.) assigns a probability weighting to the change in heart rate from x to y.

In one example, the weighting function F is a monotonically decreasing function of an absolute value of the corresponding heart rate change value |ΔHR|. One example is shown in FIGS. 10 and 11, in which the weighting function F is a monotonically decreasing exponential function of the absolute value of the corresponding heart rate change value |ΔHR|. An example is the weighting function $F(X) = \alpha * e^{(-\beta X)}$ for positive constants (α, β). In this regard, when a patient 120 is in motion, the change in heart rate on a second by second basis can be modelled as an exponentially distributed random variable $\Delta HR_{n-1 \rightarrow n}$ of the form $\text{Prob}(\Delta HR_{n-1 \rightarrow n} = X) = \alpha * e^{(-\beta X)}$, where X is the heart rate change between the previous time window n-1 and the current time window n and (α, β) are constants. In some instances, the constants (α, β) further depend on whether X is positive or negative as shown in FIGS. 10 and 11. In this regard, the function may not be bilaterally symmetrical with respect to 0 bpm. In addition, α and β may also be a function of the heart rate, water content in an individual, external temperature or other factors, and the system processor 108 can adapt these parameters or be programmed with different parameters for the weighting function. In some examples, the weighting function constants may also depend on the type of activity being performed. The processor is configured in certain examples to adapt the bandwidth of a heart rate search space according to patient activity, such as by selectively decreasing β if the patient is transitioning between low and high activity states and by selectively increasing β if the patient's activity state is constant. For instance, if an individual is transitioning from a stationary activity such as sitting down or standing to a higher intensity activity state such as running, the value of β may be smaller allowing for a larger probability weighting for a larger heart rate change. Alternatively, if the person is performing a similar activity for a longer duration of time, then the value of β may be made larger allowing a lower probability of heart change. Activity classification may be done with accelerometer in certain implementations. Note that β in the above example, is a variable that controls the bandwidth of the heart rate change search space, and can be adapted by the processor 108 for intelligent heart rate estimation according to the patient activity.

In some examples, the processor 108 implements a weighting function F of the form $F(X) = mX + c$, where c is a positive constant and m is a negative constant. In other examples F(X) may take the form of a multi-step indicator function according to the following form:

$$F(X) = \begin{cases} 1 & |X| < W11 \\ \beta_1 & W11 \leq |X| < W22 \\ \beta_2 & W22 \leq |X| \end{cases}$$

where β_2 is less than β_1 . At 418 in FIG. 4B, the processor 108 computes an updated path metric for each hypothesis at the current time window n by choosing the transition among all the heart rate hypotheses at the previous time window n-1 whose path metric+branch metric is largest. For example, the set $\{x_1, x_2, \dots, x_k\}$ denotes heart rate states from the previous time window n-1 which are within the distance threshold W1 from a potential heart rate state y1 of the current time window n. In this case, the processor 108 selects the potential path metric for the current time window state y1 according to the following formula:

$$PM(y_1, n) = \max \begin{cases} PM(x_1) + BM_{x_1 \rightarrow y_1} \\ PM(x_2) + BM_{x_2 \rightarrow y_1} \\ \vdots \\ PM(x_k) + BM_{x_k \rightarrow y_1} \end{cases}$$

In this manner, for each heart rate hypothesis HRHYP (k, n) for the current time window n, the processor 108 computes multiple individual candidate path metric values $PM(x_k, n)$ as a sum of a corresponding branch metric value $BM_{x \rightarrow y}$ and the corresponding path metric value $PM(x_k, n)$ for the corresponding heart rate hypothesis HRHYP (k, n-1) for the previous time window n-1. Moreover, for each heart rate hypothesis HRHYP (k, n) for the current time window n, the processor 108 selects the maximum one of the plurality of individual candidate path metric values $PM(x_k, n)$.

At 420 in FIG. 4B, with the path metrics updated, the processor 108 then determines the heart rate estimate value 221 as the estimate or representation the heart rate of the patient 120 for the current time window n as the heart rate hypothesis HRHYP (k, n) having a maximum candidate path metric value $PM(x_k, n)$, for example, according to the following formula: $HRE(n) = y_j = \arg \max \{PM(y_1, n), PM(y_2, n), \dots, PM(y_{N1}, n)\}$. In some embodiments, if the heart rate estimate is significantly different from the previous estimates it may be incorporated on an extra smoothing filter whose inputs are the current heart rate estimate and previous heart rate estimates. The smoothing filter may be a low-pass filter or a median filter.

At 422, the processor 108 stores (e.g., probably gates) an integer number N of the unselected states for the next iteration, and thus preserves the state of the monitored signal rate in the memory 210 (FIG. 2) for use in the next iteration time window.

At 424 in FIG. 4B, the processor 108 in one example selectively discards one or more states for the current time window n having a low path metric value $PM(x_k, n)$, e.g., below a second predetermined threshold value TH2. In this manner, the processor 108 keeps only the top k candidates (with respect to path metric) and other states are discarded. In another example, states may also be removed at 424 for the current sample time window if their path metric is less than the maximum path metric minus a third predetermined threshold value TH3.

At 426, the processor 108 may also implement selective state merging. For example, where two heart rate states are

within a predetermined distance threshold value W2 of each other for the current time window n, the processor 108 selectively merges two or more states for the current time window n. Thereafter, the process 400 returns to 402 in FIG. 4A for the next time window iteration. Thus, the system 100 provides a heart rate estimate value 221 in each time window based on the current measured frequency domain peaks as well as the history of more than one peak from the previous time window or windows as maintained by the state of the path metric values PM.

Returning to FIG. 9, the system 100 at approximately 40 seconds interprets the relatively similar heart rate frequency domain peaks of approximately 73, 91 and 184 beats per minute according to the path metric that takes the history of the PPG signal into account, and preferentially selects the illustrated true patient heart rate tracking curve 902, even if the much higher heart rate peak is significantly higher than that of the true estimate, since this estimate aligns more closely with the prior history of the patient's heart rate based on the path metric values from the previous time window. As described above, the system 100 constructs the path metric as a function of the dominant spectral amplitudes and the rate with which they change over time. The method 400 and the system 100 in certain examples can also take into account the human dynamics of heart rate change based on activity type, for example, by allowing programming and/or adaptive modification of the weighting function coefficients and constants. The system 100 in certain examples also provides a low complexity solution, e.g., for consumer products to provide a heart rate estimate to a user (e.g., sports watches, wearable health indicator devices, etc.) for a reduced complexity implementation where the spectral response for only a discrete set of frequencies needs be computed.

The above examples are merely illustrative of several possible embodiments of various aspects of the present disclosure, wherein equivalent alterations and/or modifications will occur to others skilled in the art upon reading and understanding this specification and the annexed drawings. Modifications are possible in the described embodiments, and other embodiments are possible, within the scope of the claims.

The following is claimed:

1. A heart rate monitor system to estimate a heart rate of a patient, the heart rate monitor system comprising:

a light source configured to direct a first light signal toward a patient;

an optical sensor configured to: detect a second light signal from the patient, the second light signal representing transmission or reflection of the first light signal in the patient; and generate an analog photoplethysmogram (PPG) signal representing the second light signal;

an analog to digital converter configured to generate a plurality of digital PPG sample values of the PPG signal in a time window; and

a processor circuit configured to:

for the time window: obtain a frequency domain spectrum according to the digital PPG sample values; and identify a plurality of heart rate hypothesis states having amplitudes in the frequency domain spectrum greater than a threshold value;

identify a plurality of potential heart rate transition pairs, each representing a respective heart rate transition between: a corresponding one of the heart rate hypothesis states for the time window; and one of a plurality of heart rate hypothesis states for a previous time window;

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for each potential heart rate transition pair, compute: a respective heart rate change value; and a respective branch metric value as a function of the respective heart rate change value and a frequency domain amplitude of the potential heart rate transition pair's
5 corresponding one of the heart rate hypothesis states for the time window;

for each heart rate hypothesis state for the time window: compute a plurality of candidate path metric values, each as a sum of a corresponding branch metric value and a path metric value for a corresponding one of the heart rate hypothesis states for the previous time window; and select a largest one of the candidate path metric values to be the heart rate hypothesis state's respective path metric value; and
10 determine a heart rate estimate value, representing the heart rate of the patient for the time window, as the heart rate hypothesis state whose respective path metric value is largest.

2. The heart rate monitor system of claim 1, wherein the function is a monotonically decreasing function of an absolute value of the respective heart rate change value.

3. The heart rate monitor system of claim 2, wherein the function is a monotonically decreasing exponential function of the absolute value of the respective heart rate change value.

4. The heart rate monitor system of claim 3, wherein the function is a monotonically decreasing exponential function of the form $\alpha * e^{(-\beta X)}$, where X is a heart rate change between the previous time window and the time window, α is a constant, and wherein the processor circuit is configured to adapt a bandwidth of a heart rate search space by selectively decreasing β if the patient is transitioning between low and high activity states and by selectively increasing β if the patient has a constant activity state.

5. The heart rate monitor system of claim 1, wherein the processor circuit is further configured to:

filter the digital PPG sample values to generate a plurality of filtered values corresponding to the time window; perform motion compensation processing on the filtered values to generate a plurality of motion compensated values corresponding to the time window; and
40 compute frequency content of the motion compensated values using a fast Fourier transform algorithm to generate the frequency domain spectrum for the time window.

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6. The heart rate monitor system of claim 1, wherein the processor circuit is further configured to:

if no heart rate hypothesis state for the time window is within a predetermined distance threshold value of a previous heart rate estimate for the previous time window:

determine the heart rate estimate value, representing the heart rate of the patient for the time window, as the previous heart rate estimate for the previous time window, and

propagate at least some of the heart rate hypothesis states from the previous time window as the heart rate hypothesis states for the time window.

7. The heart rate monitor system of claim 1, wherein the processor circuit is further configured to:

selectively discard potential heart rate transition pairs corresponding to heart rate change values that are greater than a predetermined distance threshold value; and

compute the respective branch metric value for each remaining potential heart rate transition pair.

8. The heart rate monitor system of claim 1, wherein the processor circuit is further configured to selectively discard one or more heart rate hypothesis states for the time window having a low path metric value.

9. The heart rate monitor system of claim 8, wherein the threshold value is a first threshold value, and the processor circuit is further configured to:

selectively discard one or more heart rate hypothesis states for the time window having a respective path metric value less than a second threshold value; and selectively discard one or more heart rate hypothesis states for the time window having a respective path metric value less than a difference between a largest one of the respective path metric values and a third threshold value.

10. The heart rate monitor system of claim 1, wherein the processor circuit is further configured to selectively merge two or more heart rate hypothesis states for the time window that are within a predetermined distance threshold value of each other.

* * * * *

专利名称(译)	具有状态序列优化的心率估计装置		
公开(公告)号	US10321831	公开(公告)日	2019-06-18
申请号	US15/237941	申请日	2016-08-16
[标]申请(专利权)人(译)	德州仪器公司		
申请(专利权)人(译)	得克萨斯仪器公司		
当前申请(专利权)人(译)	得克萨斯仪器公司		
[标]发明人	PANDE TARKESH MAGEE DAVID PATRICK NARASIMHA RAJAN		
发明人	PANDE, TARKESH MAGEE, DAVID PATRICK NARASIMHA, RAJAN		
IPC分类号	A61B5/00 A61B5/024		
CPC分类号	A61B5/02427 A61B5/02416 A61B5/7225 A61B5/7278 A61B5/7275		
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其他公开文献	US20170143218A1		
外部链接	Espacenet		

摘要(译)

所公开的示例包括心率监测系统和估计患者心率或另一脉冲信号的速率的方法，其中根据当前采样时间的脉冲信号的数字样本值，针对当前时间窗口识别速率假设或状态。窗口和速率变化值是针对当前时间窗口和先前时间窗口的状态之间的潜在速率转换而计算的。根据速率变化值和当前时间窗口的相应速率假设的频域幅度来计算转换对分支度量值，并且根据根据分支计算的最大路径度量值来确定脉冲信号速率估计。度量值和前一时间窗口的相应路径度量值。

