



(11) **EP 1 722 673 B1**

(12) **EUROPEAN PATENT SPECIFICATION**

(45) Date of publication and mention of the grant of the patent:
22.04.2009 Bulletin 2009/17

(21) Application number: **05723886.7**

(22) Date of filing: **25.02.2005**

(51) Int Cl.:
A61B 5/00 (2006.01)

(86) International application number:
PCT/US2005/006207

(87) International publication number:
WO 2005/082239 (09.09.2005 Gazette 2005/36)

(54) **SWITCH-MODE OXIMETER LED DRIVE WITH A SINGLE INDUCTOR**

SCHALTMODUS-OXIMETER-LED-ANTRIEB MIT EINEM EINZIGEN INDUKTOR

ATTAQUE DE DEL DE SPHYGMO-OXYMETRE A MODE DE COMMUTATION AVEC UN SEUL INDUCTEUR

(84) Designated Contracting States:
AT BE BG CH CY CZ DE DK EE ES FI FR GB GR HU IE IS IT LI LT LU MC NL PL PT RO SE SI SK TR

(30) Priority: **25.02.2004 US 787852**

(43) Date of publication of application:
22.11.2006 Bulletin 2006/47

(73) Proprietor: **Nellcor Puritan Bennett LLC**
North Haven, CT 06473 (US)

(72) Inventors:
• **CHEW, Bradford, B.**
San Ramon, CA 94583 (US)

• **PETERSEN, Ethan**
Castro Valley, CA 94546 (US)
• **SHEA, William**
Livermore, CA 94550 (US)

(74) Representative: **Grubert, Andreas**
Baker Botts (UK) LLP
41 Lothbury
London EC2R 7HF (GB)

(56) References cited:
GB-A- 2 044 550 **US-A- 5 165 162**
US-B1- 6 198 405 **US-B1- 6 665 551**

EP 1 722 673 B1

Note: Within nine months of the publication of the mention of the grant of the European patent in the European Patent Bulletin, any person may give notice to the European Patent Office of opposition to that patent, in accordance with the Implementing Regulations. Notice of opposition shall not be deemed to have been filed until the opposition fee has been paid. (Art. 99(1) European Patent Convention).

Description

[0001] The present invention relates to oximeters according to the pre-amble of claim 1, and in particular to LED drive circuits in pulse oximeters. A driver circuit and method of A current driving system of a light emitting diode is known from US 6,665,551 B1. Operating the same is known from US 6198405.

[0002] Pulse oximetry is typically used to measure various blood chemistry characteristics including, but not limited to, the blood-oxygen saturation of hemoglobin in arterial blood, the volume of individual blood pulsations supplying the tissue, and the rate of blood pulsations corresponding to each heartbeat of a patient. Measurement of these characteristics has been accomplished by use of a non-invasive sensor which scatters light through a portion of the patient's tissue where blood perfuses the tissue, and photoelectrically senses the absorption of light in such tissue. The amount of light absorbed at various wavelengths is then used to calculate the amount of blood constituent being measured.

[0003] The light scattered through the tissue is selected to be of one or more wavelengths that are absorbed by the blood in an amount representative of the amount of the blood constituent present in the blood. The amount of transmitted light scattered through the tissue will vary in accordance with the changing amount of blood constituent in the tissue and the related light absorption. For measuring blood oxygen level, such sensors have typically been provided with a light source that is adapted to generate light of at least two different wavelengths, and with photodetectors sensitive to both of those wavelengths, in accordance with known techniques for measuring blood oxygen saturation.

[0004] Known non-invasive sensors include devices that are secured to a portion of the body, such as a finger, an ear or the scalp. In animals and humans, the tissue of these body portions is perfused with blood and the tissue surface is readily accessible to the sensor.

[0005] The light sources, typically light emitting diodes (LEDs), need to be driven with current to activate them. Because of the significant amount of current required, this can interfere with reducing power consumed by an oximeter. One solution is shown in U.S. Patent No. 6,226,539. There, an inductor and capacitor circuit is used to first store charge in a first switch position, and then subsequently, in a second switch position, deliver that stored charge to the LED. Two different inductor and capacitor circuits are used, one for each LED. It would be desirable to reduce the number of components required in the circuit of this patent.

[0006] It is an object of the present invention to provide a light emitter drive circuit for an oximeter. This object can be achieved by the features of the independent claim 1. Further enhancements are characterized in the dependent claims.

[0007] The present invention relates to a light emitter drive circuit according to claim 1. The present invention

provides a light emitter drive circuit for an oximeter which utilizes a single inductor for driving multiple light emitters. The inductor is connected through a switching circuit to multiple energy storage circuits, such as capacitors. These are alternately charged up, using the same inductor. Subsequently, the capacitors are alternately discharged to activate their corresponding light emitters through the same inductor.

[0008] In another aspect of the present invention, the magnetic susceptibility of the LED drive circuit is reduced by using magnetic flux canceling in the inductor. In one embodiment, a toroidal inductor is used with geometric symmetry in its magnetic flux. In another embodiment, a dual core closed bobbin shielded inductor is used. This embodiment has windings of both cores in series that are used to cancel the effect of an external magnetic field.

[0009] For a further understanding of the nature and advantages of the present invention, reference should be made to the following description taken in conjunction with the accompanying drawings.

[0010] Fig. 1 is a block diagram of an oximeter incorporating the present invention.

[0011] Fig. 2 is a circuit diagram of a LED drive circuit according to an embodiment of the present invention.

[0012] Fig. 3 is a block diagram of one embodiment of the logic for generating the timing and control signals for the circuit of Fig. 2.

[0013] Fig. 4 is a diagram of a toroidal inductor used in one embodiment of the present invention.

[0014] Figs. 5 and 6 are diagrams of a dual core inductor according to an embodiment of the present invention.

Oximeter system

[0015] Fig. 1 illustrates an embodiment of an oximetry system incorporating the present invention. A sensor 10 includes red and infrared LEDs and a photodetector. These are connected by a cable 12 to a board 14. LED drive current is provided by an LED drive interface 16. The received photocurrent from the sensor is provided to an I-V interface 18. The IR and red voltages are then provided to a sigma-delta interface 20 incorporating the present invention. The output of sigma-delta interface 20 is provided to a microcontroller 22. Microcontroller 22 includes flash memory for a program, and SRAM memory for data. The processor also includes a microprocessor chip 24 connected to a flash memory 26. Finally, a clock 28 is used and an interface 30 to a digital calibration in the sensor 10 is provided. A separate host 32 receives the processed information, as well as receiving an analog signal on a line 34 for providing an analog display. LED Drive Circuit

[0016] Fig. 2 is a circuit diagram of the LED drive circuit according to an embodiment of the invention, which forms a portion of LED drive interface 16 of Fig. 1. A voltage regulator 36 provides a voltage separate from the voltage supply for the overall oximeter circuitry. The

output is provided as a 4.5 volt signal on line 38, with the level being set by the feedback resistor divider of resistors R89 and R90. The voltage on line 38 is provided to a FET transistor Q11 to an inductor L6. The current through inductor L6 is provided by a switch 40 to one of capacitors C65 and C66, which store charge for the red and IR LEDs, respectively. A red/IR control signal on line 42 selects the switch position under control of the oximeter processor. A control signal LED PWM gate on line 44 controls the switching of transistor switch Q11.

[0017] Once the capacitors are charged up, the control signal on line 44 turns off switch Q11 and current is provided from either capacitor C65 or C66, through switch 40 and inductor L6 to either the red anode line 46 or the IR anode line 48 by way of transistors Q5 and Q6, respectively. A signal "red gate" turns on transistor Q5, while its inverse, "red gate" turns off transistor Q7. This provides current through the red anode line 46 to the back to back LEDs 50, with the current returning through the IR anode to transistor Q8 and through resistor R10 to ground. Transistor Q8 is turned on by the signal "IR gate" while the inverse of this signal, "IR gate" turns off transistor Q6. The signals are reversed when the IR anode is to be driven, with the "IR gate" and "red gate" signals, and their inverses, changing state, so that current is provided through transistor Q6 to IR anode 48 and returns through red anode 46 and through transistor Q7 to resistor R10 and ground. The "LED current sense" signal is read for calibration purposes not relevant to the present invention.

[0018] When the current from the capacitor C65 or C66 is provided through inductor L6 to the LEDs, and that current is switched off at the desired time, transistor Q11 is turned on so that the remaining current during the transition can be dumped into capacitor C64. This addresses the fact that the FET transistor switching is not instantaneous. Subsequently, C64 will dump its current through Q11 and inductor L6 into the capacitors when they are recharged.

[0019] Resistor R38 and capacitor C67 are connected in parallel to inductor L6 to protect against signal spikes, and provide a smooth transition. Connected to inductor L6 is a sampling circuit with a switch 52 controlled by an LED sample hold signal on line 54 to sample the signals and provide them through an amplifier 56 to a "LED current" signal on line 58 which is read by the processor. Operational amplifier 56 operates between 4.5 volts and ground. Thus, a voltage reference slightly above ground, of 0.2 volts, is provided as a voltage reference on pin 3. An integrating capacitor C68 is provided in parallel to amplifier 56. A switch 60 responds to a "clear LED sample" signal to operate the switch to short out the capacitor between samples.

[0020] The sample and hold circuit measures the voltage at node T18, between capacitor C69 and inductor L6, to determine the current. Capacitor C69 is 1/1000 of the value of capacitors C65 and C66. Thus, a proportional current is provided through C69, which is injected through

switch 52 to integrating capacitor C68 to provide a voltage which can be measured at the output of amplifier 56 on line 58. The voltage measured by the processor on line 58 is used as a feedback, with the processor varying the width of the pulse delivered to transistor Q11 to selectively vary the amount of energy that's delivered to the capacitors 65 and 66, and then is eventually discharged to the LEDs 50. A PI (Proportional Integral) loop inside the processor then controls the PWM signal at Q11. This allows precise control of the LED intensity, allowing it to be maximized, if desired, without exceeding the desired limits (to avoid burning the patient, etc.).

[0021] The lower left of the diagram shows a "4.5 v LED disable" signal which is used by the microprocessor to turn off the voltage regulator 36 in certain instances. For example, diagnostics looking for shorts in a new sensor plugged in will turn off the voltage regulator if there is a problem with the LED line.

[0022] Fig. 3 illustrates processor 22, from Fig. 1, connected to programmed logic 62, which is in the LED drive interface 16 in Fig. 1. Programmed logic 62 provides the different control signals used by the circuit of Fig. 2 in response to basic timing signals from the processor of a clock, a sync pulse, and a pulse width signal.

[0023] Thus, the present invention provides an improvement over the circuit shown in Patent No. 6,226,539 by moving the switch position between the inductor and the capacitors to eliminate the need for two inductors. This not only reduces the part count, requiring only one inductor instead of two, but also provides better matching between the red and IR drive currents since both use the same inductor.

[0024] In another aspect of the invention, the LED drive circuit's susceptibility to magnetic interference is reduced. This magnetic interference can distort the detected pleth waveform. This is minimized by using magnetic flux canceling in the inductor. In one embodiment, this is a toroidal inductor as shown in Fig. 4. The toroidal inductor has a geometric symmetry in its magnetic flux. Another embodiment uses a dual core closed bobbin shielded inductor, such as shown in Figs. 5 and 6. The windings of both cores in series are used to cancel the effect of an external magnetic field. These magnetic flux canceling inductors can be used either in the circuit of Fig. 2, or could be used in the dual inductor embodiment of the prior art. Fig. 5 shows the dual core inductor with a bobbin 70 in a cylinder 72. The wires are wound through gaps 76, as shown in Fig. 6. A first winding 78 is clockwise, while a second winding 80 is counterclockwise. A top view 82 is also shown. Ideally, the combined inductance in one embodiment is 680 uH.

[0025] The invention as illustrated in the embodiment of Fig. 2 enables the multiplexing of current, through an H-bridge topology, to back-to-back LEDs. Alternately, a different number of loads could be provided. The present invention is scalable to N-loads. The present invention provides significant efficiencies through reduction of support components, choice of components, and the prop-

erties of "loss-less" capacitor and inductor storage devices. The circuit of Fig. 2 can handle a range of forward voltage drops across the LEDs. The voltage provided varies automatically in accordance with the LED voltage drop, and does not put out more energy than it needs to.

[0026] The circuit is dynamically controlled through a PI loop in the processor, with current feedback being provided by the capacitive current divider from each storage capacitor (C65 and C66), which provides isolation. The feedback can be calibrated with a traditional in-line sense resistor, RIO. In addition, this technique allows adjustment of the peak current for optimal signal-to-noise during the sampling period.

[0027] The addition of the upstream linear regulator 36 enhances power supply rejection capability, while the PI loop provides additional power supply insensitivity (to drift, P-P, surge, etc.).

[0028] As will be appreciated by those with skill in the art, the present invention can be embodied in other specific forms without departing from the essential characteristics thereof. For example, instead of two drive lines, three drive lines could be provided by adding another leg with FET transistor switches connected to the inductor. Additionally, this could be scalable to more than three legs connected in parallel, similar to the leg of Q6, Q8, and the leg of Q5, Q7.

Claims

1. A light emitter drive circuit for an oximeter comprising:

a voltage generating circuit;
 an inductor (L6) having a first node coupled to said voltage generating circuit through a switch (Q11); and
 a first switching circuit (40) coupled to a second node of said inductor (L6) wherein the switch (Q11) is not part of the first switching circuit;

characterized by further comprising first and second capacitors (C65, C66) connected to said first switching circuit (40), such that current from said inductor (L6) is fed through said first switching circuit (40) steering said current to respectively said first or second capacitor (C65, C66) when said first switching circuit (40) is in a respective first or second selected switch position, each of said capacitor (C65, C66) being capable of storing energy to drive a light emitter;
 a second switching circuit (Q5, Q6, Q7, Q8) different from said first switching circuit coupled to the first node of said inductor (L6); and
 first and second light emitter drive lines connected to said second switching circuit, such that stored energy from each of said first and second capacitors (C65, C66) can be coupled to said first and second

light emitter drive lines for driving a said light emitter.

2. The circuit of claim 1, wherein said drive lines are configured to drive light emitters that are back-to-back.
3. The circuit of claim 1 or 2, wherein said second switching circuit is an H-bridge drive circuit including four transistors (Q5, Q6, Q7, Q8).
4. The circuit according to one of the preceding claims, wherein said voltage generating circuit is a voltage regulator (36) separate from a supply voltage for a remainder of said oximeter.
5. The circuit according to one of the preceding claims, further comprising a resistor (R38) and a capacitor (C67) in parallel with said inductor (L6) for driving a said light emitter.
6. The circuit according to one of the preceding claims, wherein said inductor (L6) is a magnetic flux canceling inductor.
7. The circuit of claim 6, wherein said inductor (L6) is a toroidal inductor.
8. The circuit of claim 6, wherein said inductor (L6) is a dual core closed bobbin shielded inductor.
9. The circuit according to one of the preceding claims, further comprising a sample and hold circuit connected to said second node of said inductor (L6).
10. The circuit according to one of the preceding claims, further comprising at least a third energy storage circuit.

Patentansprüche

1. Lichtemitteransteuerschaltung für ein Oximeter, aufweisend:

eine Spannungserzeugungsschaltung;
 einen Induktor (L6), welcher eine erste Verbindung aufweist, die mit der Spannungserzeugungsschaltung durch einen Schalter (Q11) gekoppelt ist; und
 einen ersten Schaltkreis (40), welcher mit einer zweiten Verbindung des Induktors (L6) gekoppelt ist, wobei der Schalter (Q11) nicht Teil des ersten Schaltkreises ist; **dadurch gekennzeichnet, dass** diese ferner aufweist erste und zweite Kondensatoren (C65, C66), welche mit dem ersten Schaltkreis (40) derart verbunden sind, dass von dem Induktor (L6) Strom durch den ersten Schaltkreis (40), der den

- Strom steuert, jeweils an den ersten und zweiten Kondensator (C65, C66) zugeführt wird, wenn der erste Schaltkreis (40) in einer jeweiligen ersten oder zweiten gewählten Schaltposition ist, wobei jeder Kondensator (C65, C66) in der Lage ist, Energie zu speichern, um einen Lichtemitter anzusteuern;
- einen zweiten Schaltkreis (Q5, Q6, Q7, Q8), welcher sich von dem ersten Schaltkreis unterscheidet und mit der ersten Verbindung des Induktors (L6) gekoppelt ist; und
- erste und zweite Lichtemitteransteuerleitungen, welche mit dem zweiten Schaltkreis derart verbunden sind, dass gespeicherte Energie von jedem der ersten und zweiten Kondensatoren (C65, C66) wechselweise mit den ersten und zweiten Lichtemitteransteuerleitungen zum Ansteuern eines Lichtemitters gekoppelt werden können.
2. Schaltung nach Anspruch 1, wobei die Ansteuerleitungen konfiguriert sind, Lichtemitter anzusteuern, welche aufeinander folgen.
 3. Schaltung nach Anspruch 1 oder 2, wobei der zweite Schaltkreis eine B2-Ansteuerschaltung ist, welche vier Transistoren enthält (Q5, Q6, Q7, Q8).
 4. Schaltung nach einem der vorhergehenden Ansprüche, wobei die Spannungserzeugungsschaltung ein von einer Versorgungsspannung für einen Rest des Oximeters separater Spannungsregler (36) ist.
 5. Schaltung nach einem der vorhergehenden Ansprüche, welche ferner einen Widerstand (R38) und einen Kondensator (C67) aufweisen, welche zu dem Induktor (L6) parallel verbunden sind.
 6. Schaltung nach einem der vorhergehenden Ansprüche, wobei der Induktor (L6) ein Magnetflusstilgungsinduktor ist.
 7. Schaltung nach Anspruch 6, wobei der Induktor (L6) ein ringförmiger Induktor ist.
 8. Schaltung nach Anspruch 6, wobei der Induktor (L6) ein abgeschirmter Doppelkerninduktor mit geschlossenem Spulenkörper ist.
 9. Schaltung nach einem der vorhergehenden Ansprüche, welche eine Abtast- und Halteschaltung aufweisen, welche mit der zweiten Verbindung des Induktors (L6) verbunden ist.
 10. Schaltung nach einem der vorhergehenden Ansprüche, welche zumindest eine dritte Energiespeicherschaltung aufweisen.

Revendications

1. Circuit de commande d'émetteur de lumière pour oxymètre comprenant :
 - un circuit de génération de tension;
 - une inductance (L6) possédant un premier noeud couplé audit circuit de production de tension par un commutateur (Q11) ; et
 - un premier circuit de commutation (40) couplé à un second noeud de ladite inductance (L6) dans lequel le commutateur (Q11) ne fait pas partie du premier circuit de commutation ;

caractérisé en ce qu'il comprend, en outre un premier et un second condensateurs (C65, C66) connectés audit premier circuit de commutation (40), de sorte que le courant de ladite inductance (L6) est acheminé par ledit premier circuit de commutation (40) envoyant ledit courant, respectivement, vers ledit premier ou second condensateur (C65, C66), lorsque ledit premier circuit de commutation (40) se trouve dans une première ou seconde position de commutation sélectionnée respective, ledit condensateur (C65, C66) étant capable de stocker de l'énergie pour commander un émetteur de lumière ; un second circuit de commutation (Q5, Q6, Q7, Q8) différent dudit premier circuit de commutation couplé au premier noeud de ladite inductance (L6) ; et une première et une seconde lignes de commande d'émetteur de lumière connectées audit second circuit de commutation, de sorte que l'énergie stockée par chacun desdits premier et second condensateurs (C65, C66) peut être alternativement couplée auxdites première et seconde lignes de commande d'émetteur de lumière pour commander un dit émetteur de lumière.
2. Circuit de la revendication 1, dans lequel lesdites lignes de commande sont configurées pour commander des émetteurs de lumière disposés tête-bêche.
3. Circuit de la revendication 1 ou 2, dans lequel ledit second circuit de commutation est un circuit de commande de pont en H incluant quatre transistors (Q5, Q6, Q7, Q8).
4. Circuit selon l'une des revendications précédentes, dans lequel ledit circuit de génération de tension est un régulateur de tension (36) distinct d'une tension d'alimentation pour une partie restante dudit oxymètre.
5. Circuit selon l'une des revendications précédentes, comprenant en outre une résistance (R38) et un condensateur (C67) monté en parallèle avec ladite inductance (L6).

6. Circuit selon l'une des revendications précédentes, dans lequel ladite inductance (L6) est une inductance d'annulation de flux magnétique.
7. Circuit de la revendication 6, dans lequel ladite inductance (L6) est une inductance toroïdale. 5
8. Circuit de la revendication 6, dans lequel ladite inductance (L6) est une inductance blindée bobinée fermée à double noyau. 10
9. Circuit selon l'une des revendications précédentes, comprenant en outre un circuit échantillonneur-bloqueur connecté audit second noeud de ladite inductance (L6). 15
10. Circuit selon l'une des revendications précédentes, comprenant en outre au moins un troisième circuit de stockage d'énergie. 20

25

30

35

40

45

50

55

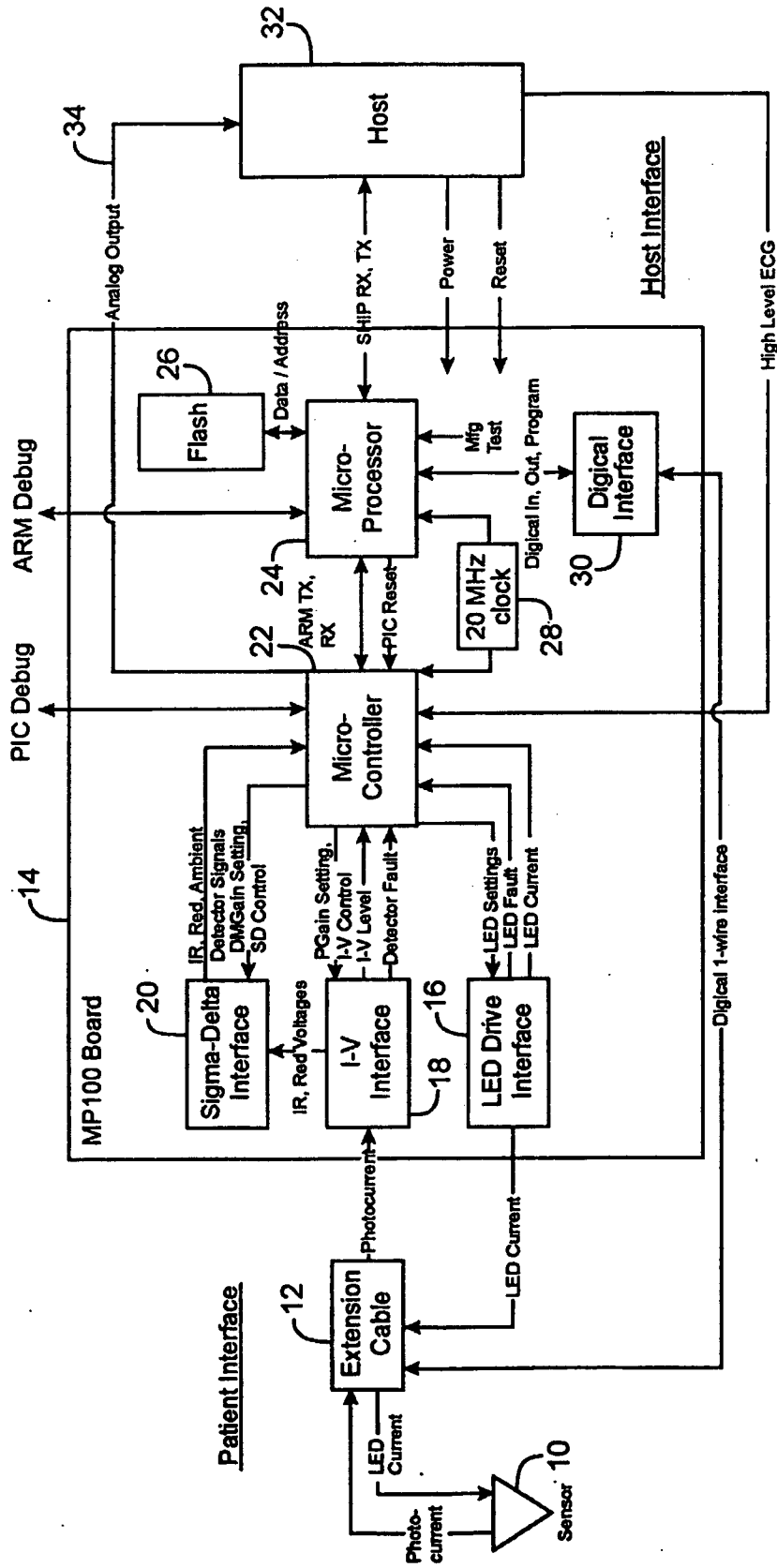


FIG. 1

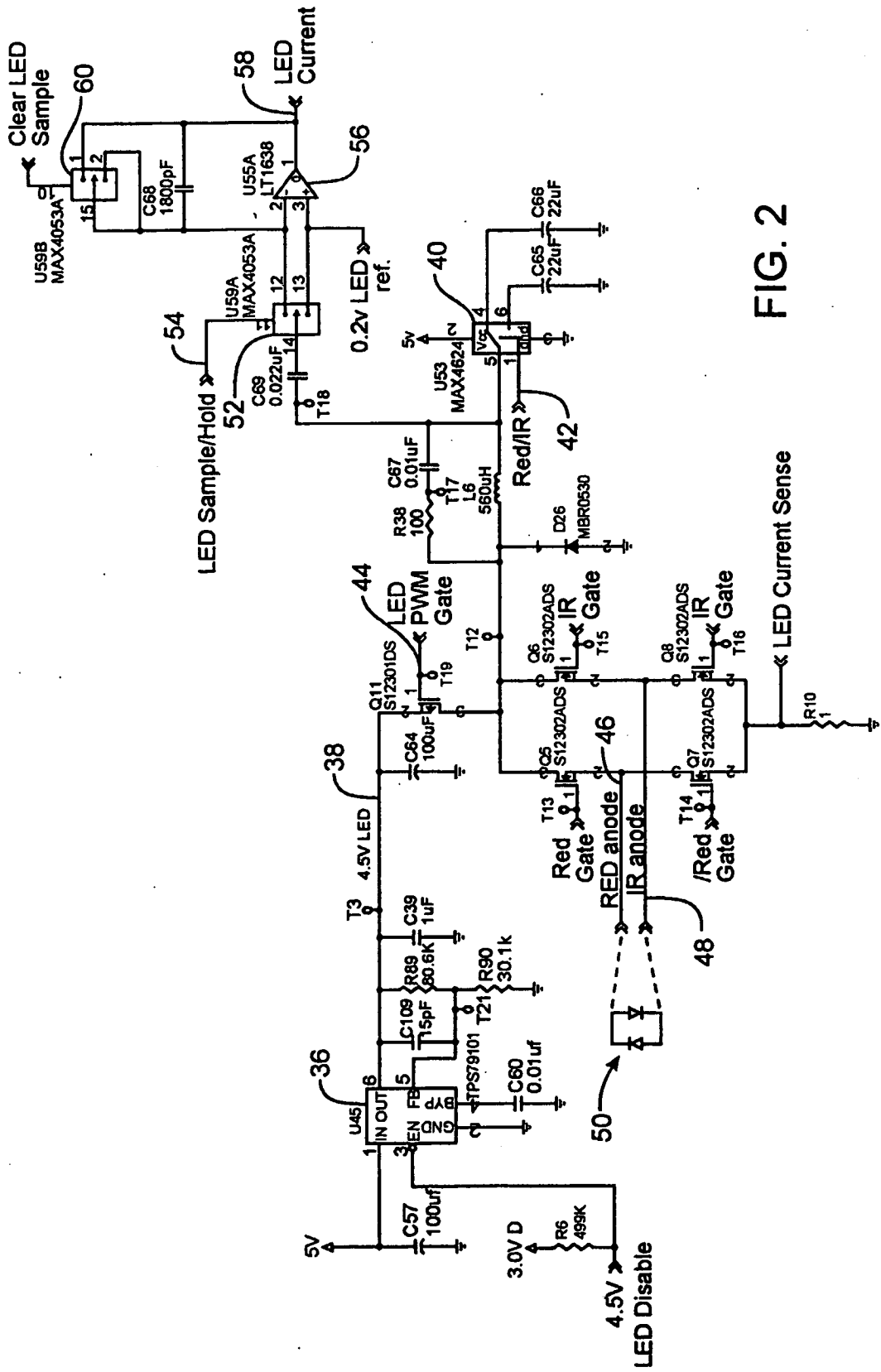


FIG. 2

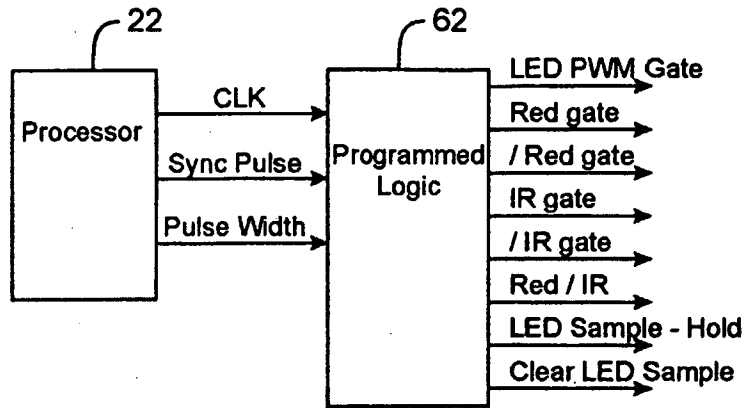


FIG. 3



FIG. 4

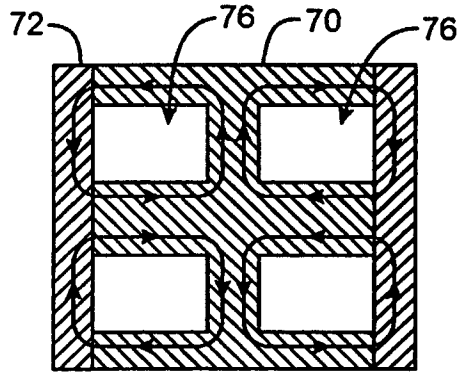


FIG. 5

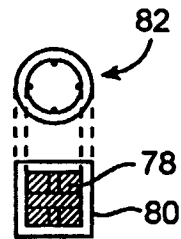


FIG. 6

REFERENCES CITED IN THE DESCRIPTION

This list of references cited by the applicant is for the reader's convenience only. It does not form part of the European patent document. Even though great care has been taken in compiling the references, errors or omissions cannot be excluded and the EPO disclaims all liability in this regard.

Patent documents cited in the description

- US 6665551 B1 [0001]
- US 6198405 B [0001]
- US 6226539 B [0005] [0023]

